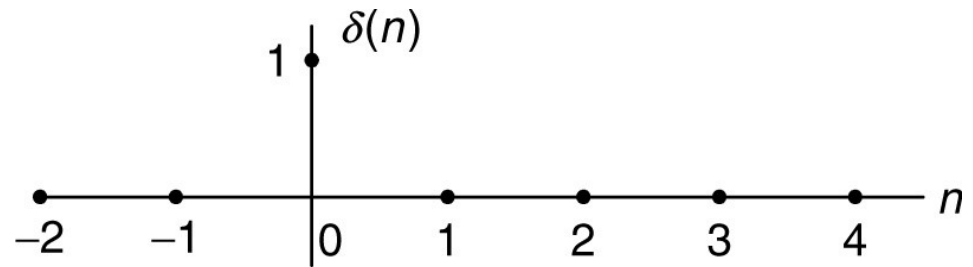


Common Digital Sequences

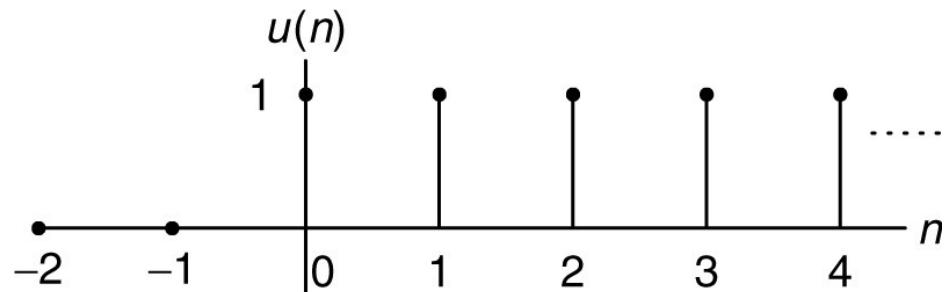
Unit-impulse sequence:

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & n \neq 0 \end{cases}$$



Unit-step sequence:

$$u(n) = \begin{cases} 1 & n \geq 0 \\ 0 & n < 0 \end{cases}$$

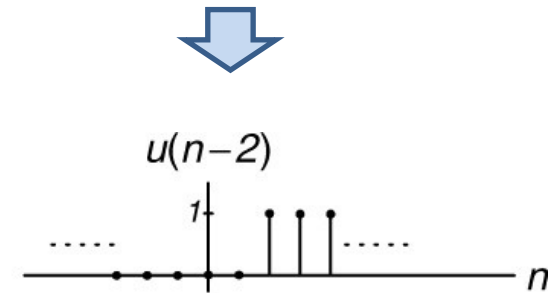
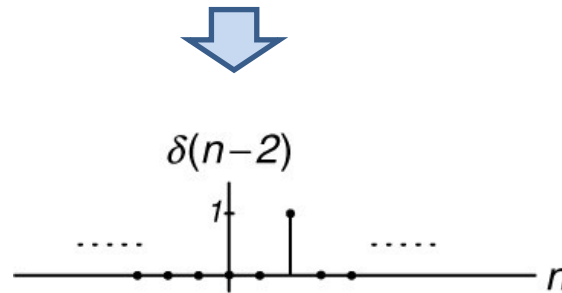


Shifted Sequences

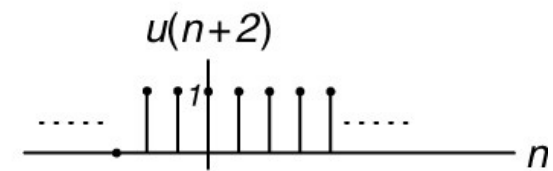
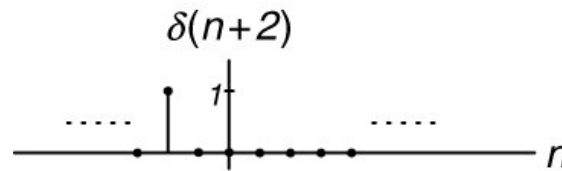
Shifted unit-impulse

Shifted unit-step

Right shift by two samples



Left shift by two samples



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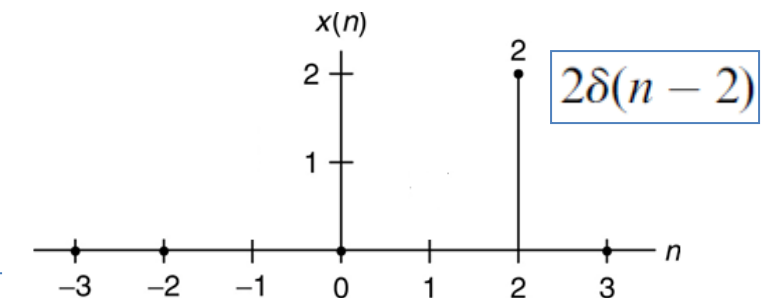
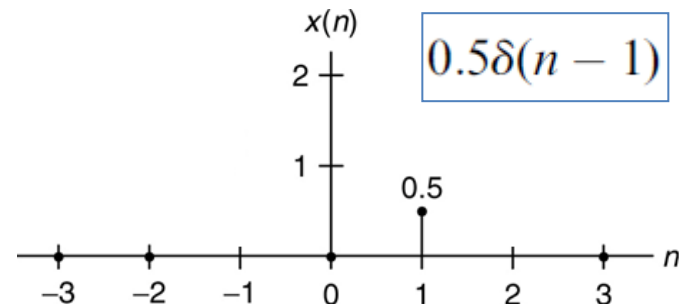
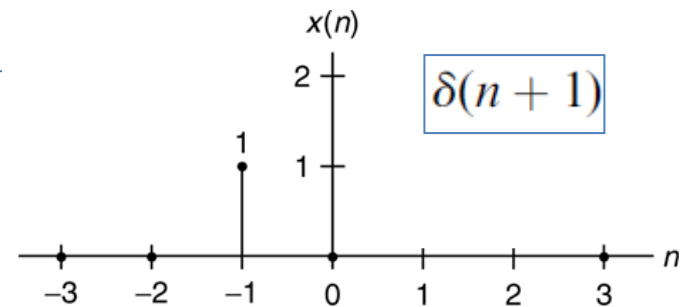
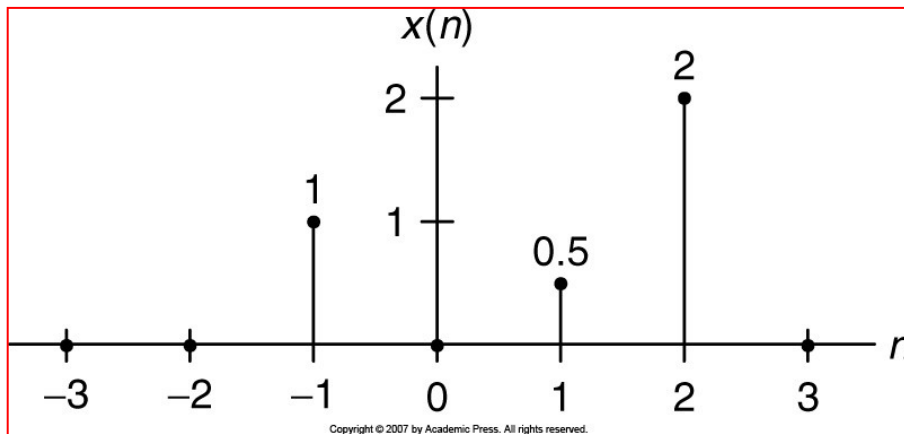
Example 1

Given the following,

$$x(n] = \delta(n + 1) + 0.5\delta(n - 1) + 2\delta(n - 2),$$

a. Sketch this sequence.

Solution:



Generation of Digital Signals

Let, sampling interval, $\Delta t = T$

$x(n)$: digital signal

$x(t)$: analog signal

$$x(n) = x(t)|_{t=nT} = x(nT)$$

Also $u(t)|_{t=nT} = u(nT) = u(n)$

Example 2

Convert analog signal $x(t)$ into digital signal $x(n)$, when sampling period is 125 microsecond, also plot sample values.

$$x(t) = 10e^{-5000t}u(t)$$

Solution:

$$t = nT = n \times 0.000125 = 0.000125n$$

$$x(n) = x(nT) = 10e^{-5000 \times 0.000125n}u(nT) = 10e^{-0.625n}u(n)$$

Example 2 (contd.)

The first five
sample values:



$$x(0) = 10e^{-0.625 \times 0} u(0) = 10.0$$

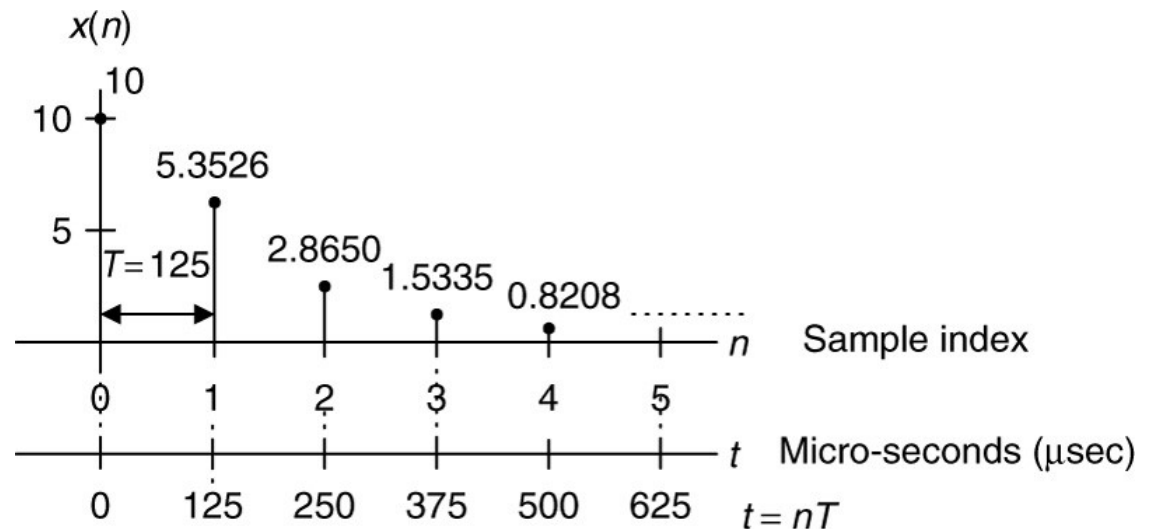
$$x(1) = 10e^{-0.625 \times 1} u(1) = 5.3526$$

$$x(2) = 10e^{-0.625 \times 2} u(2) = 2.8650$$

$$x(3) = 10e^{-0.625 \times 3} u(3) = 1.5335$$

$$x(4) = 10e^{-0.625 \times 4} u(4) = 0.8208$$

Plot of the digital
sequence:

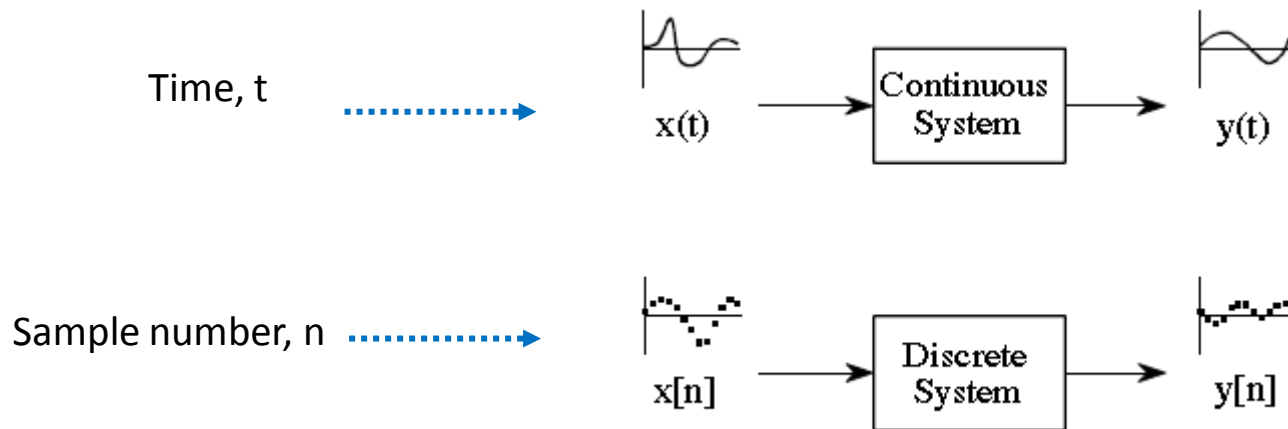


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Linear System

System: A system that produces an output signal in response to an input signal.

Continuous system & discrete system.



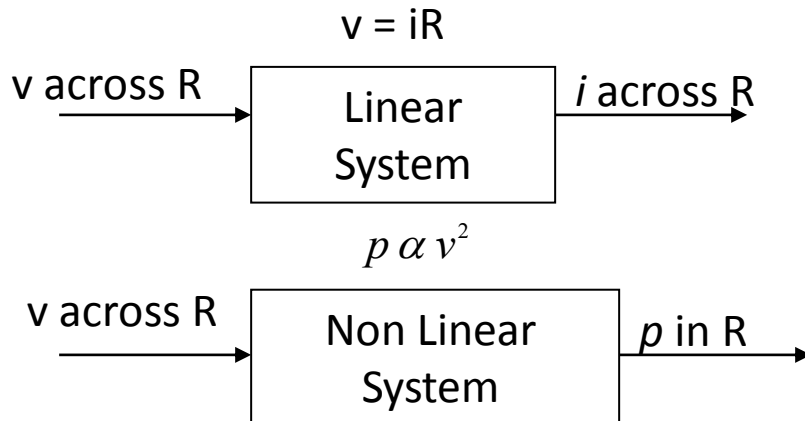
Linear Systems: Property 1

1. **Homogeneity** |
 2. **Additivity** |
- Must for all linear systems
3. **Shift invariance** → Must for DSP linear systems

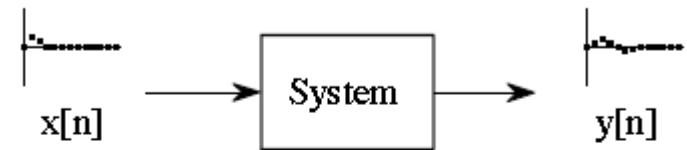
Homogeneity: (deals with amplitude)

If $x[n] \rightarrow y[n]$, then $kx[n] \rightarrow ky[n]$

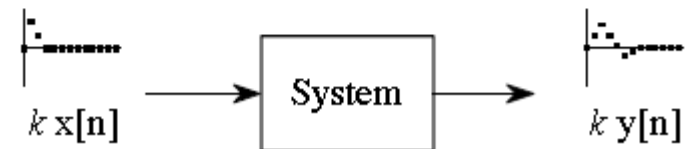
K is a constant



IF

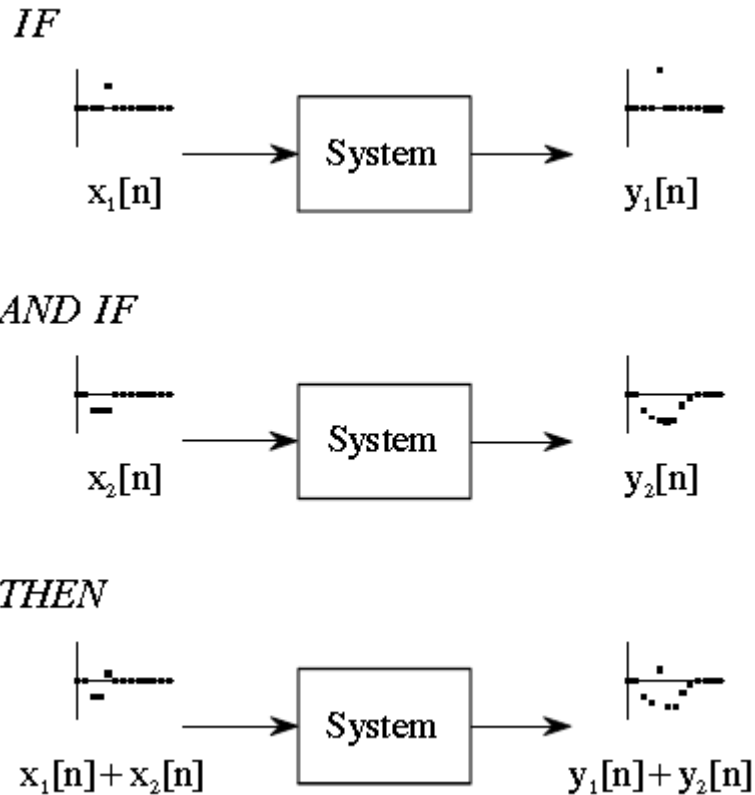


THEN

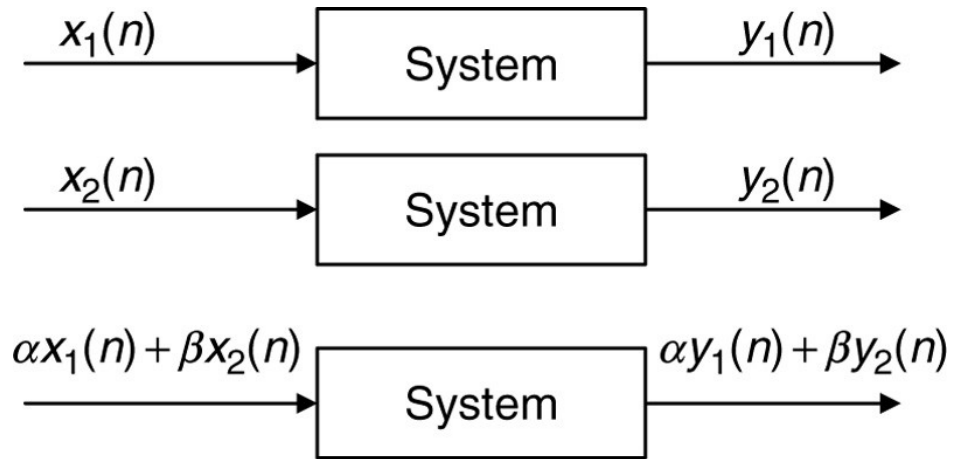


Linear Systems: Property 2

Additivity



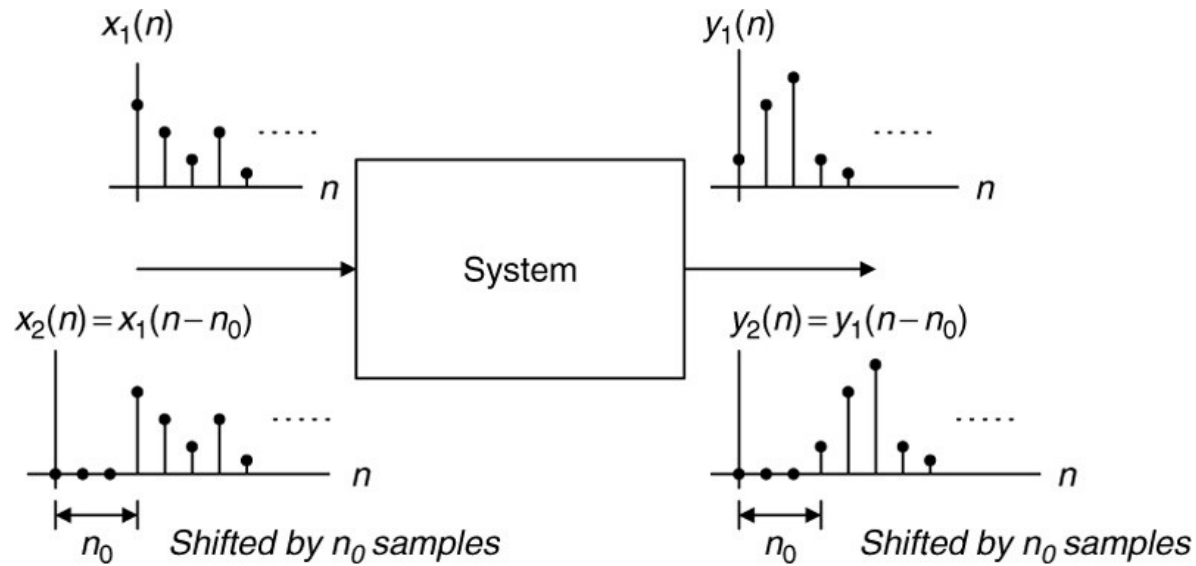
Homogeneity & Additivity



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Linear Systems: Property 3

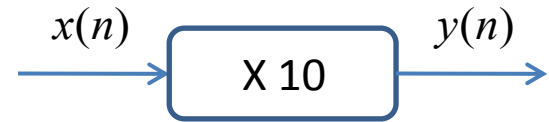
Shift (time) Invariance



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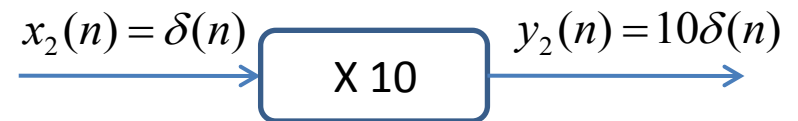
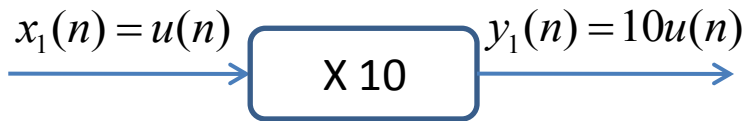
Example 3

Let a digital amplifier, $y(n) = 10x(n)$



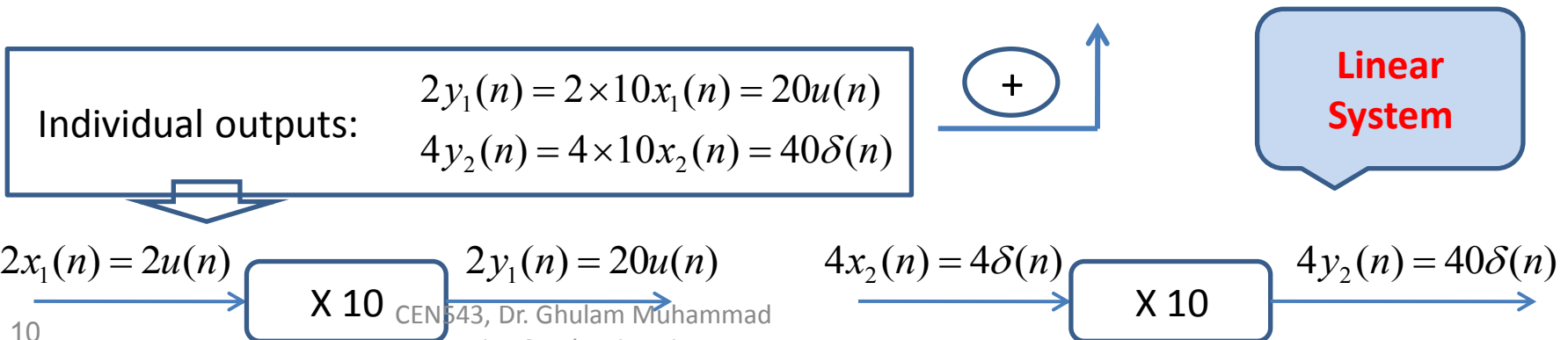
If the inputs are: $x_1(n) = u(n)$ and $x_2(n) = \delta(n)$

Outputs will be: $y_1(n) = 10u(n)$ and $y_2(n) = 10\delta(n)$, respectively.

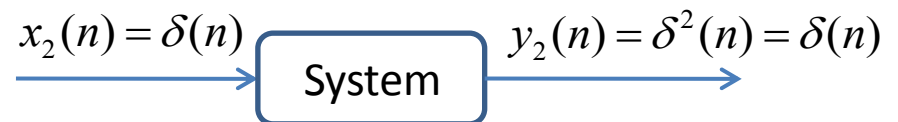
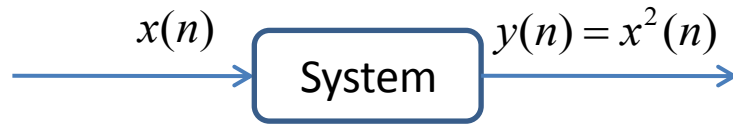


If we apply combined input to the system: $x(n) = 2x_1(n) + 4x_2(n) = 2u(n) + 4\delta(n)$

The output will be: $y(n) = 10x(n) = 10(2u(n) + 4\delta(n)) = 20u(n) + 40\delta(n)$



Example 4



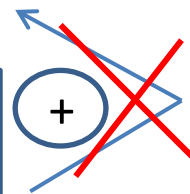
If the input is: $4x_1(n) + 2x_2(n)$

$$\begin{aligned} \text{Then the output is: } y(n) &= x^2(n) = (4x_1(n) + 2x_2(n))^2 \\ &= (4u(n) + 2\delta(n))^2 = 16u^2(n) + 16u(n)\delta(n) + 4\delta^2(n) \\ &= 16u(n) + 20\delta(n). \end{aligned}$$

Individual outputs:

$$4y_1(n) = 4 \times x_1^2(n) = 4u(n)$$

$$2y_2(n) = 2 \times x_2^2(n) = 2\delta(n)$$

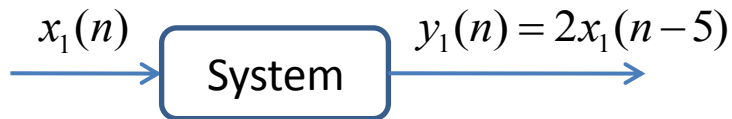


Non Linear System

Example 5 (a)

Given the linear system $y(n) = 2x(n - 5)$, find whether the system is time invariant or not.

Solution:



Let the shifted input be: $x_2(n) = x_1(n - n_0)$

Therefore system output: $y_2(n) = 2x_2(n - 5) = 2x_1(n - n_0 - 5)$. ← Equal

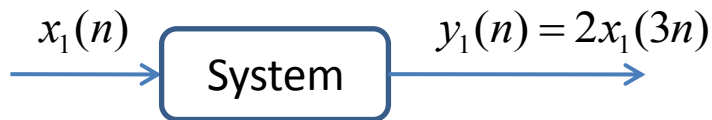
Shifting $y_1(n) = 2x_1(n - 5)$ by n_0 samples leads to $y_1(n - n_0) = 2x_1(n - 5 - n_0)$.

Time Invariant

Example 5 (b)

Given the linear system $y(n) = 2x(3n)$, find whether the system is time invariant or not.

Solution:



Let the shifted input be: $x_2(n) = x_1(n - n_0)$

Therefore system output: $y_2(n) = 2x_2(3n) = 2x_1(3n - n_0)$ ← NOT Equal

Shifting $y_1(n) = 2x_1(3n)$ by n_0 samples leads to

$$y_1(n - n_0) = 2x_1(3(n - n_0)) = 2x_1(3n - 3n_0)$$

NOT Time Invariant

Difference Equation

A causal, linear, and time invariant system can be represented by a difference equation as follows:

$$\underbrace{y(n) + a_1y(n-1) + \dots + a_Ny(n-N)}_{\text{Outputs}} = \underbrace{b_0x(n) + b_1x(n-1) + \dots + b_Mx(n-M)}_{\text{Inputs}}$$

After rearranging:

$$y(n) = -a_1y(n-1) - \dots - a_Ny(n-N) + b_0x(n) + b_1x(n-1) + \dots + b_Mx(n-M)$$

Finally:

$$y(n) = -\sum_{i=1}^N a_i y(n-i) + \sum_{j=0}^M b_j x(n-j)$$

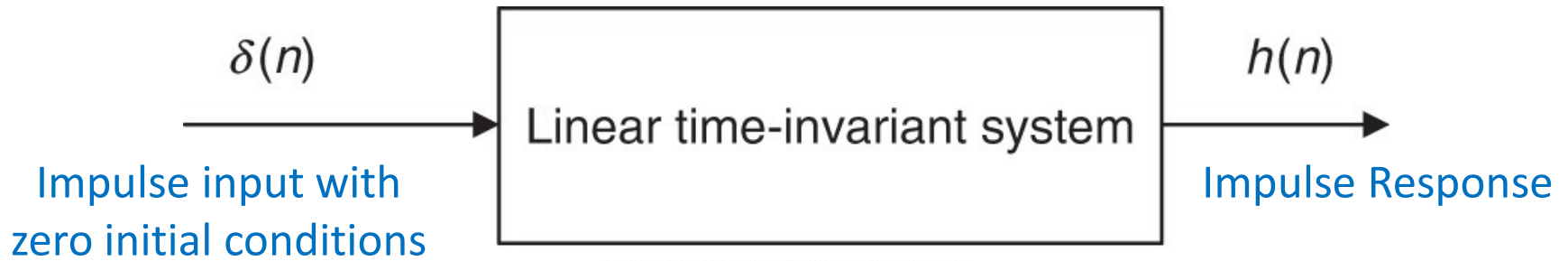
Example 6

Identify non zero system coefficients of the following difference equations.

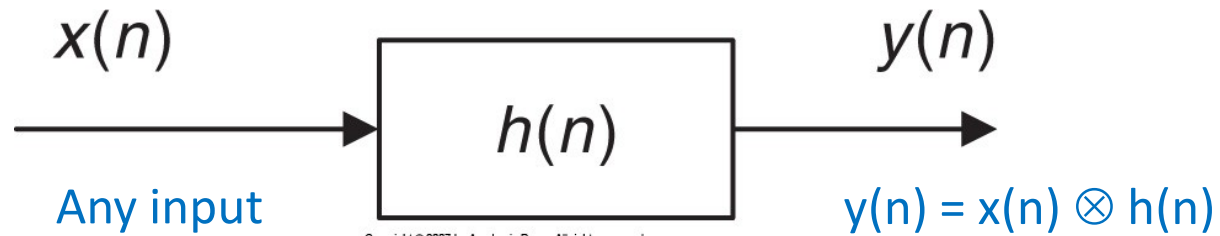
$$y(n) = 0.25y(n - 1) + x(n) \quad \xrightarrow{\text{Solution:}} \quad b_0 = 1, \quad a_1 = -0.25$$

$$y(n) = x(n) + 0.5x(n - 1) \quad \xrightarrow{\text{Solution:}} \quad b_0 = 1, \quad b_1 = 0.5$$

System Representation Using Impulse Response



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Convolution

$$y(n) = \dots + h(-1)x(n+1) + h(0)x(n) + h(1)x(n-1) + h(2)x(n-2) + \dots$$

Example 7 (a)

Given the linear time-invariant system:

$$y(n] = 0.5x(n] + 0.25x(n - 1) \text{ with an initial condition } x(-1) = 0,$$

- Determine the unit-impulse response $h(n)$.
- Draw the system block diagram.
- Write the output using the obtained impulse response.

Solution:

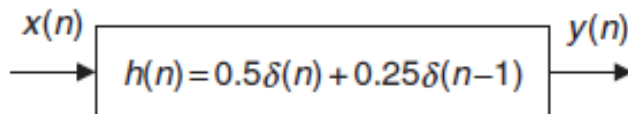
a. let $x(n] = \delta(n]$

$$h(n] = y(n] = 0.5x(n] + 0.25x(n - 1) = 0.5\delta(n] + 0.25\delta(n - 1)$$

Therefore,

$$h(n] = \begin{cases} 0.5 & n = 0 \\ 0.25 & n = 1 \\ 0 & \text{elsewhere} \end{cases}$$

b.



c. $y(n] = h(0)x(n] + h(1)x(n - 1)$

Example 7 (b)

Given the difference equation

$$y(n] = 0.25y(n - 1) + x(n) \text{ for } n \geq 0 \text{ and } y(-1) = 0,$$

- Determine the unit-impulse response $h(n)$.
- Draw the system block diagram.
- Write the output using the obtained impulse response.

Solution:

a. let $x(n] = \delta(n)$ Then $h(n] = 0.25h(n - 1) + \delta(n)$.

$$h(0] = 0.25h(-1) + \delta(0] = 0.25 \times 0 + 1 = 1$$

$$h(1] = 0.25h(0] + \delta(1] = 0.25 \times 1 + 0 = 0.25$$

$$h(2] = 0.25h(1] + \delta(2] = 0.25 \times 0.25 + 0 = 0.0625$$

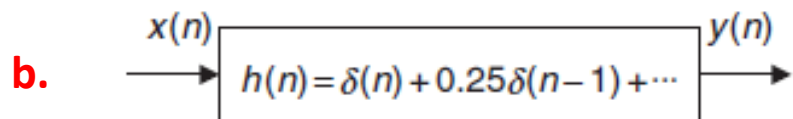
...

With the calculated results, we can predict the impulse response as

$$h(n] = (0.25)^n u(n] = \delta(n] + 0.25\delta(n - 1) + 0.0625\delta(n - 2) + \dots$$

 Infinite!

Example 7 (b) - contd.



c.
$$y(n) = h(0)x(n) + h(1)x(n-1) + h(2)x(n-2) + \dots$$
$$= x(n) + 0.25x(n-1) + 0.0625x(n-2) + \dots$$

Finite Impulse Response (FIR) system:

When the difference equation contains no previous outputs, i.e. 'a' coefficients are zero. < See example 7 (a) >

Infinite Impulse Response (IIR) system:

When the difference equation contains previous outputs, i.e. 'a' coefficients are not all zero. < See example 7 (b) >

BIBO Stability

BIBO: Bounded In and Bounded Out

A stable system is one for which every bounded input produces a bounded output.

$$y(n) = \dots + h(-1)x(n+1) + h(0)x(n) + h(1)x(n-1) + h(2)x(n-2) + \dots$$

Let, in the worst case, every input value reaches to maximum value M .

$$y(n) = M(\dots + h(-1) + h(0) + h(1) + h(2) + \dots).$$

Using absolute values of the impulse responses,

$$y(n) < M(\dots + |h(-1)| + |h(0)| + |h(1)| + |h(2)| + \dots).$$

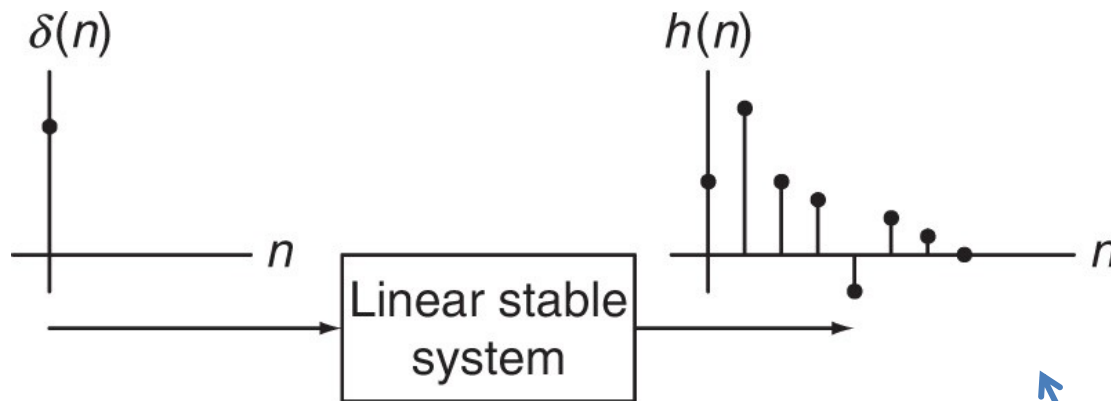
If the impulse responses are finite number, then **output is also finite.**

Stable system.

BIBO Stability - contd.

To determine whether a system is stable, we apply the following equation:

$$S = \sum_{k=-\infty}^{\infty} |h(k)| = \dots + |h(-1)| + |h(0)| + |h(1)| + \dots < \infty.$$



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Impulse response is decreasing to zero.

Example 8

Given a linear system given by: $y(n) = 0.25y(n - 1) + x(n)$ for $n \geq 0$ and $y(-1) = 0$

Which is described by the unit-impulse response: $h(n) = (0.25)^n u(n)$

Determine whether the system is stable or not.

Solution:

$$S = \sum_{k=-\infty}^{\infty} |h(k)| = \sum_{k=-\infty}^{\infty} |(0.25)^k u(k)|$$

Using definition of step function:

$$u(k) = 1 \text{ for } k \geq 0.$$



$$S = \sum_{k=0}^{\infty} (0.25)^k = 1 + 0.25 + 0.25^2 + \dots$$

For $a < 1$, we know $\sum_{k=0}^{\infty} a^k = \frac{1}{1-a}$ where $a = 0.25 < 1$

$$\text{Therefore } S = 1 + 0.25 + 0.25^2 + \dots = \frac{1}{1-0.25} = \frac{4}{3} < \infty$$

The summation is finite, so the system is stable.

Digital Convolution

$$y(n) = \sum_{k=-\infty}^{\infty} h(k)x(n-k)$$

$$= \dots + h(-1)x(n+1) + h(0)x(n) + h(1)x(n-1) + h(2)x(n-2) + \dots$$

The sequences are interchangeable.



Commutative

$$y(n) = \sum_{k=-\infty}^{\infty} h(k)x(n-k) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$

$$x[n]*h[n] = h[n]*x[n]$$

Convolution sum requires $h(n)$ to be reversed and shifted.

If $h(n)$ is the given sequence, $h(-n)$ is the reversed sequence.

Reversed Sequence

Given a sequence,

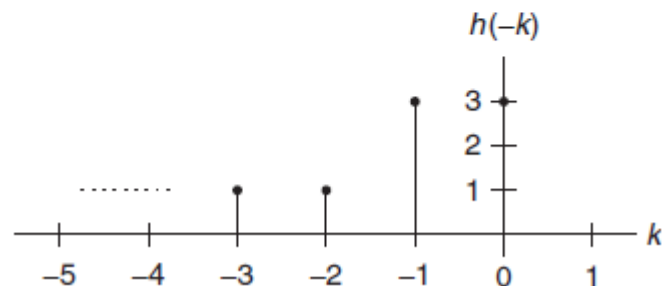
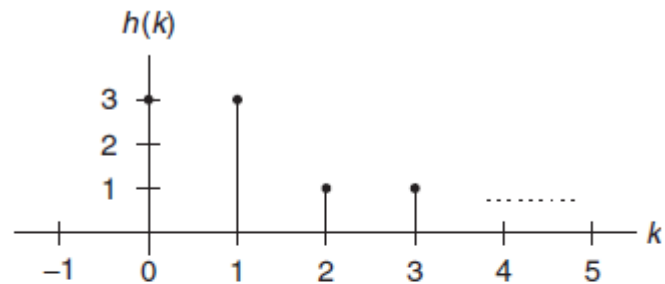
$$h(k) = \begin{cases} 3, & k = 0,1 \\ 1, & k = 2,3 \\ 0 & \text{elsewhere} \end{cases}$$

where k is the time index or sample number,

- a. Sketch the sequence $h(k)$ and reversed sequence $h(-k)$.

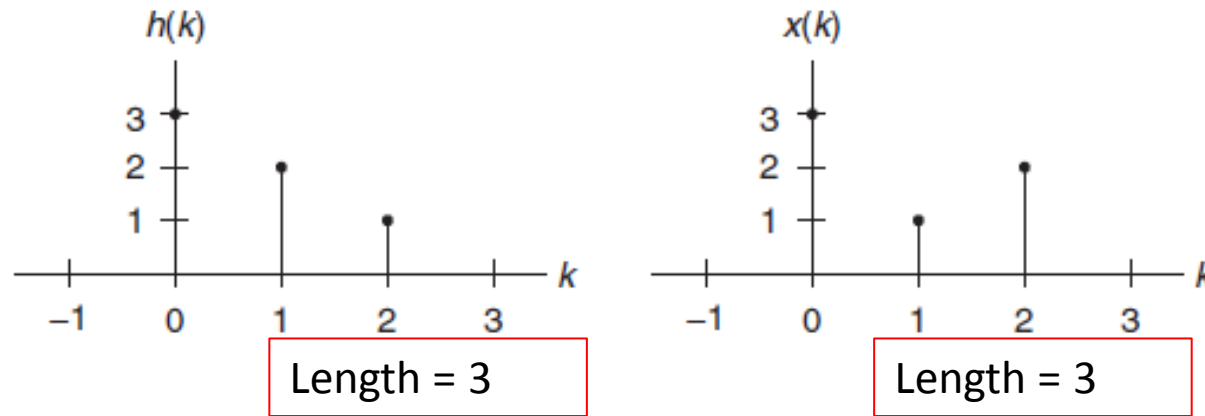
Solution:

a.



Convolution Using Table Method

Example 9



Solution:

Convolution sum using the table method.

$k:$	-2	-1	0	1	2	3	4	5	
$x(k):$			3	1	2				
$h(-k):$	1	2	3						$y(0) = 3 \times 3 = 9$
$h(1-k)$		1	2	3					$y(1) = 3 \times 2 + 1 \times 3 = 9$
$h(2-k)$			1	2	3				$y(2) = 3 \times 1 + 1 \times 2 + 2 \times 3 = 11$
$h(3-k)$				1	2	3			$y(3) = 1 \times 1 + 2 \times 2 = 5$
$h(4-k)$					1	2	3		$y(4) = 2 \times 1 = 2$
$h(5-k)$						1	2	3	$y(5) = 0$ (no overlap)

Convolution length = $3 + 3 - 1 = 5$

Convolution Using Table Method

Example 10

$$x(n) = \begin{cases} 1 & n = 0, 1, 2 \\ 0 & \text{otherwise} \end{cases} \text{ and } h(n) = \begin{cases} 0 & n = 0 \\ 1 & n = 1, 2 \\ 0 & \text{otherwise} \end{cases}$$

Length = 3

Length = 2

Solution:

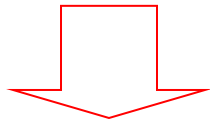
$k:$	-2	-1	0	1	2	3	4	5	...	
$x(k):$			1	1	1				...	
$h(-k):$	1	1	0							$y(0) = 0$ (no overlap)
$h(1-k)$		1	1	0						$y(1) = 1 \times 1 = 1$
$h(2-k)$			1	1	0					$y(2) = 1 \times 1 + 1 \times 1 = 2$
$h(3-k)$				1	1	0				$y(3) = 1 \times 1 + 1 \times 1 = 2$
$h(4-k)$					1	1	0			$y(4) = 1 \times 1 = 1$
$h(n-k)$						1	1	0		$y(n) = 0, n \geq 5$ (no overlap)
										Stop

Convolution length = $3 + 2 - 1 = 4$

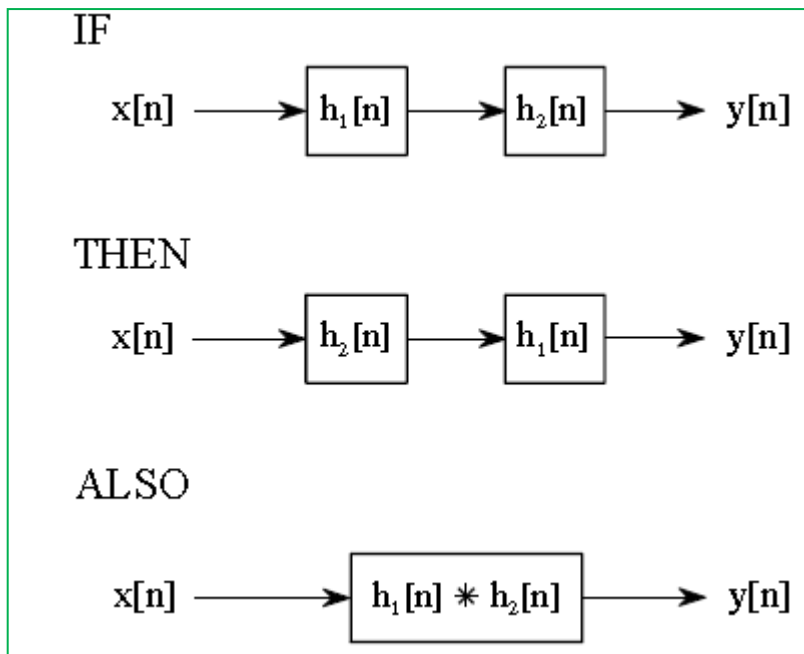
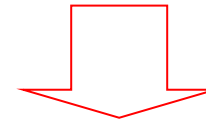
Convolution Properties

Commutative: $a[n] * b[n] = b[n] * a[n]$

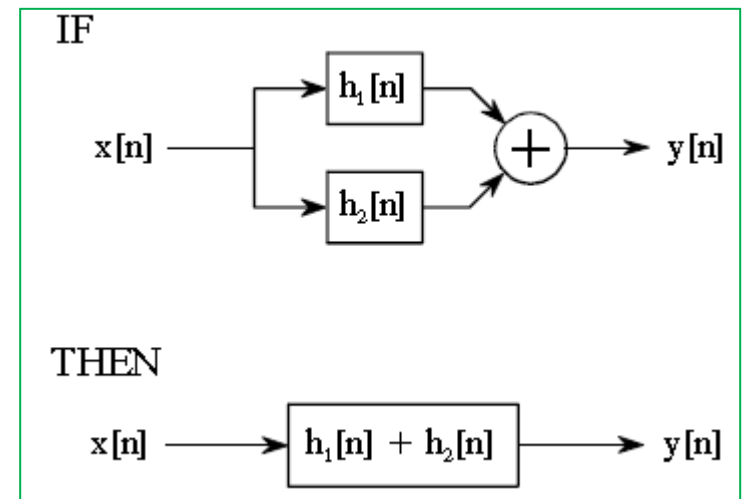
Associative: $(a[n] * b[n]) * c[n] = a[n] * (b[n] * c[n])$



Distributive: $a[n] * b[n] + a[n] * c[n] = a[n] * (b[n] + c[n])$



Associative

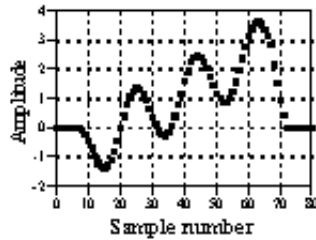


Distributive

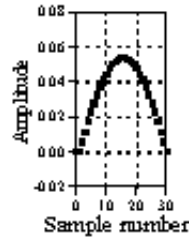
Examples of Convolution

Kernel

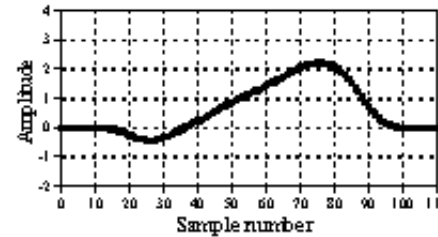
a. Low-pass Filter



*

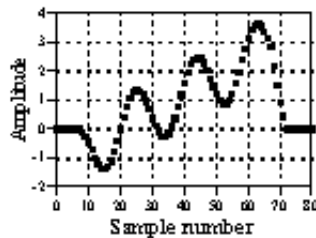


=

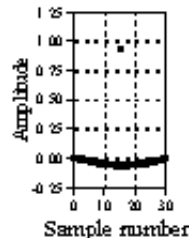


Slow rising ramp

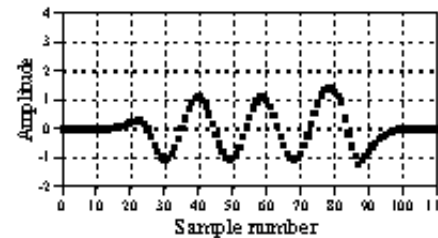
b. High-pass Filter



*



=



High freq. Sine wave

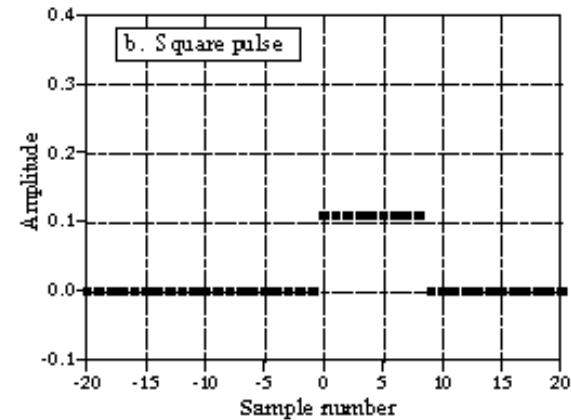
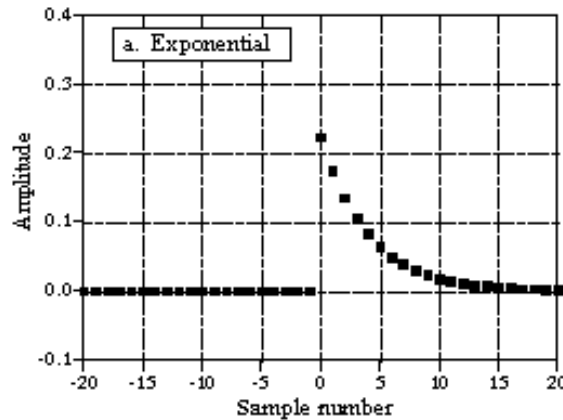
Input Signal

Impulse Response

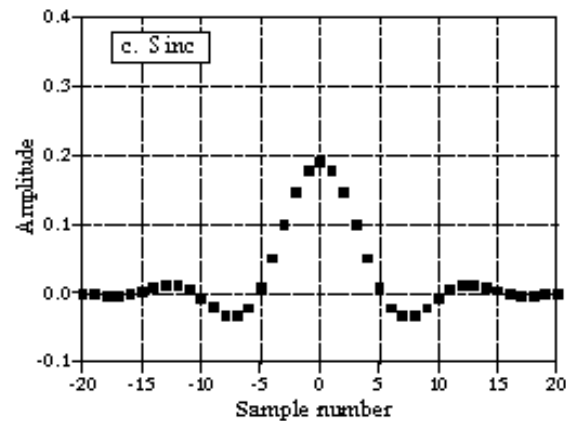
Output Signal

Low Pass Filters

Kernel: formed by a group of positive adjacent points that provide smoothing.



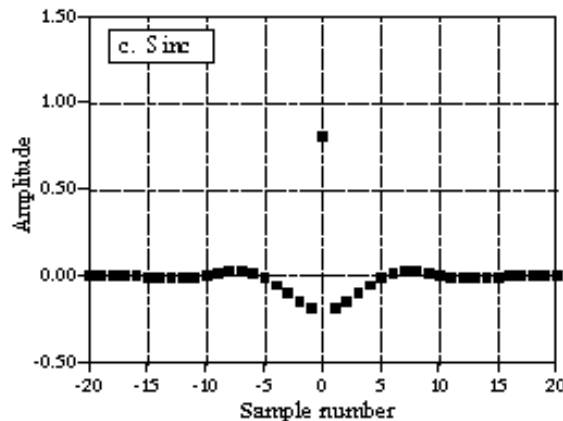
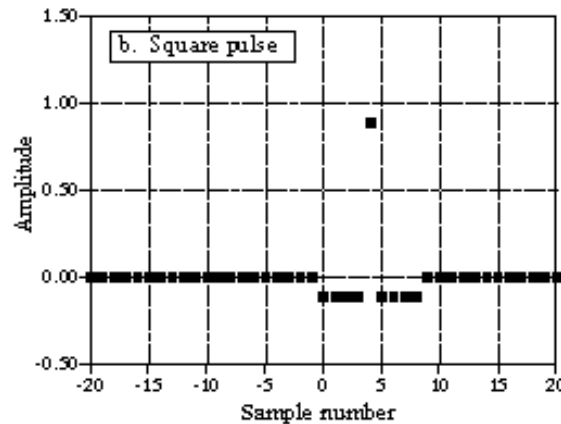
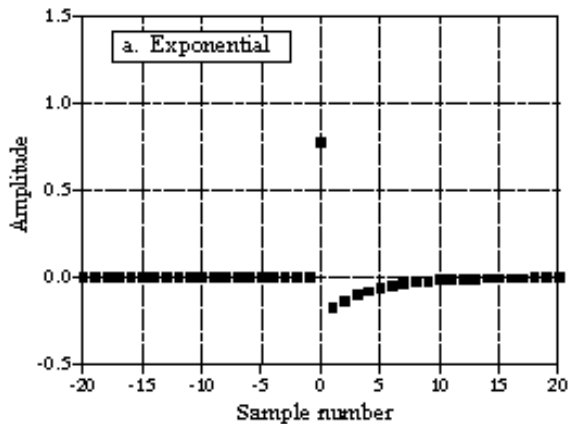
For
reducing
noise



Sum of the points must be one. Gain of *one* in DC.

High Pass Filters

Kernel: delta function – corresponding low-pass filter.



Peak is surrounded by many adjacent negative points.

Sum of the points must be zero. Zero gain at DC (zero frequency).

Signal-to-Noise Ratio (SNR)

Bel or decibel (dB):

A bel: The power is changed by a *factor of ten*.

3 bels \rightarrow Power of $10 \times 10 \times 10 = 1000$ times.

Decibel (dB): One-tenth of a bel.

30 dB \rightarrow Power of $10 \times 10 \times 10 = 1000$ times.

$$dB = 10 \log_{10} \frac{P_2}{P_1}$$

$$dB = 20 \log_{10} \frac{A_2}{A_1}$$

0 dB $\rightarrow 10^0$ times = 1 time = equal power.

Clean signal, $s(n)$, with variance = 0.5

Noise signal, $v(n)$, with variance = 1

$$\sigma_x^2 = \sigma_s^2 + K^2 \sigma_v^2$$

Noisy signal, $x(n) = s(n) + Kv(n)$

Find K so that SNR = 20 dB.

$$SNR = 20dB = 10 \log_{10} \left(\frac{\text{var}(s(n))}{\text{var}(v(n))} \right) = 10 \log_{10} \left(\frac{0.5}{K^2} \right)$$

$$\log_{10} \frac{0.5}{K^2} = 2 \Rightarrow \frac{0.5}{K^2} = 10^2 \Rightarrow K = 0.07071$$

Periodicity

Example 11

Consider the following continuous signal for the current $i(t) = \cos(20\pi t)$ which is sampled at 12.5 ms. Will the resulting discrete signal be periodic?

The continuous radian frequency is $\omega = 20\pi$ radians. Since the sampling rate interval $T_s = 12.5 \text{ msec} = 0.0125 \text{ sec}$, then

$$x(n) = \cos(2\pi(10)(0.0175)n) = \cos\left(\frac{2\pi}{8}n\right) = \cos\left(\frac{\pi}{4}n\right)$$

Since for periodicity we must have: $\frac{2\pi}{\theta_0} = \frac{N}{k}$

We get, $\frac{2\pi}{2\pi/8} = \frac{N}{k} = \frac{16\pi}{2\pi} = \frac{8}{1}$

For $k = 1$ we have $N = 8$, which is the fundamental period.

If N/k is a rational number (ratio of two integers) then $x(n)$ is periodic and the period is

$$N = k \left(\frac{2\pi}{\theta_0} \right)$$

The smallest value of N that satisfies the above equation is called the fundamental period. If $2\pi/\theta_0$ is not a rational number, then $x(n)$ is not periodic.

Figure Acknowledgement

Most of the figures are taken from the following books:

Li Tan, *Digital Signal Processing, Fundamentals and Applications*, Elsevier, 2008.

Steven W. Smith, *Digital Signal Processing: A Practical Guide for Engineers and Scientists*, Newnes, Elsevier, 2003.