
Logic and Computer Design Fundamentals

Chapter 7 – Registers and Register Transfers

Part 2 – Counters, Register Cells, Buses, & Serial Operations

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Overview

- **Part 1 – Registers, Microoperations and Implementations**
- **Part 2 – Counters, register cells, buses, & serial operations**
 - **Microoperations on single register (continued)**
 - **Counters**
 - **Register cell design**
 - **Multiplexer and bus-based transfers for multiple registers**
 - **Serial transfers and microoperations**
- **Part 3 – Control of Register Transfers**

Counters

- **Counters** are sequential circuits which "count" through a specific state sequence. They can **count up**, **count down**, or **count through other fixed sequences**. Two distinct types are in common usage:

1. **Ripple Counters**

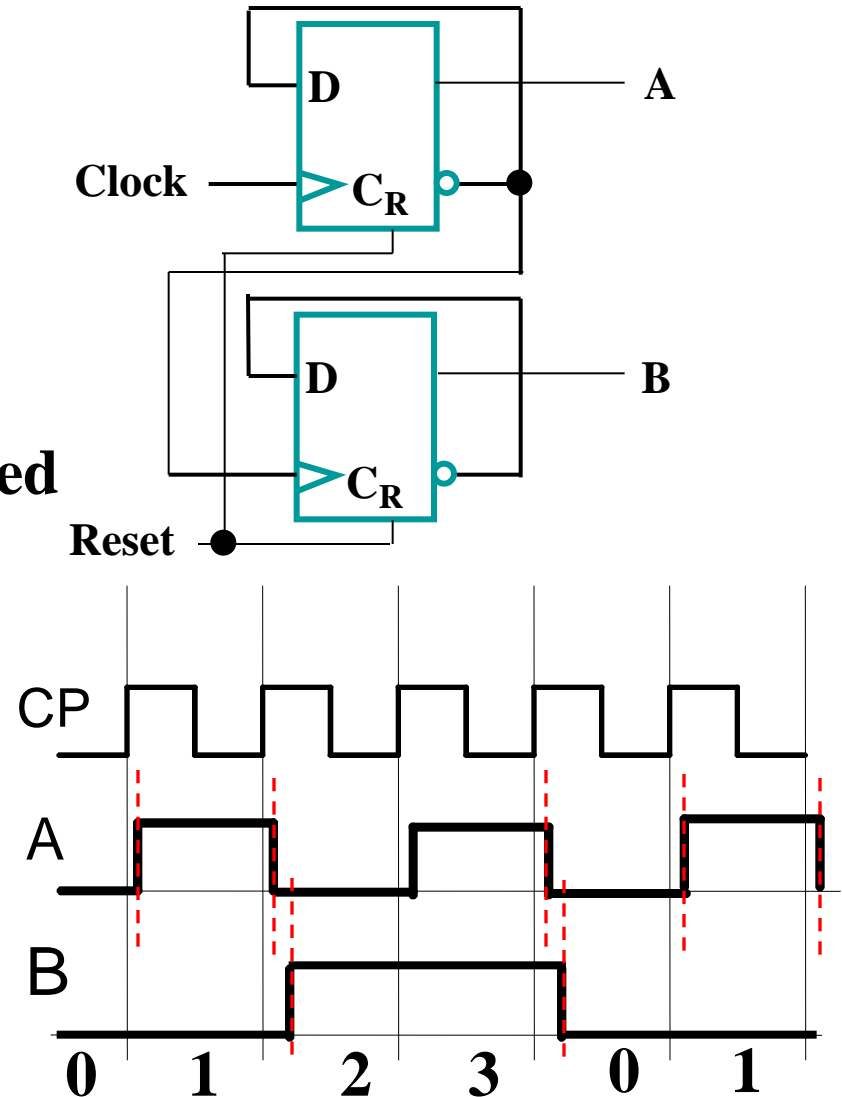
- Clock connected to the flip-flop clock input on the LSB bit flip-flop
- For all other bits, a flip-flop output is connected to the clock input, thus circuit is not truly synchronous!
- Output change is delayed more for each bit toward the MSB.
- Resurgent because of low power consumption

2. **Synchronous Counters**

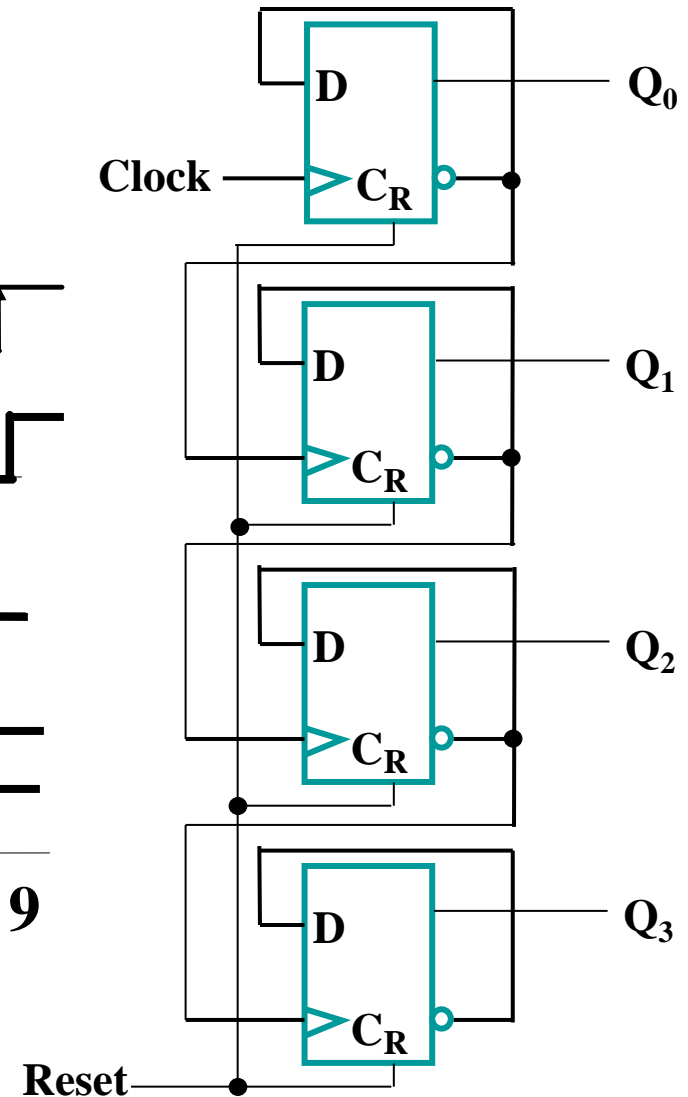
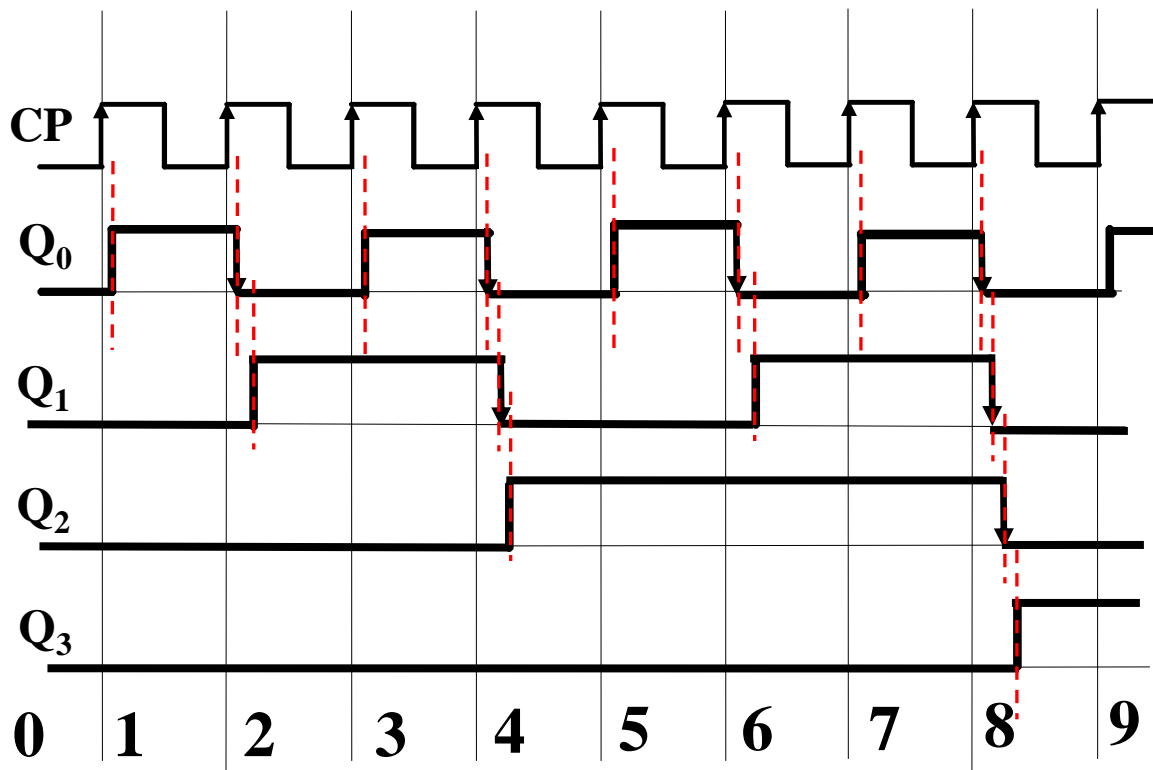
- Clock is directly connected to the flip-flop clock inputs
- Logic is used to implement the desired state sequencing

Ripple Counter

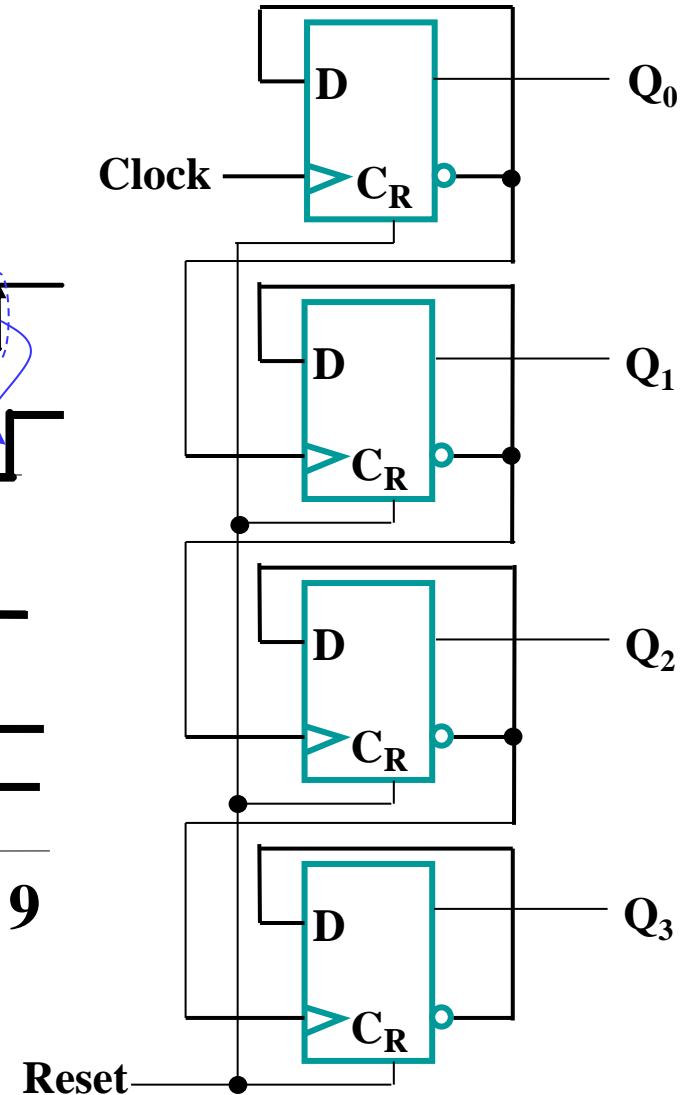
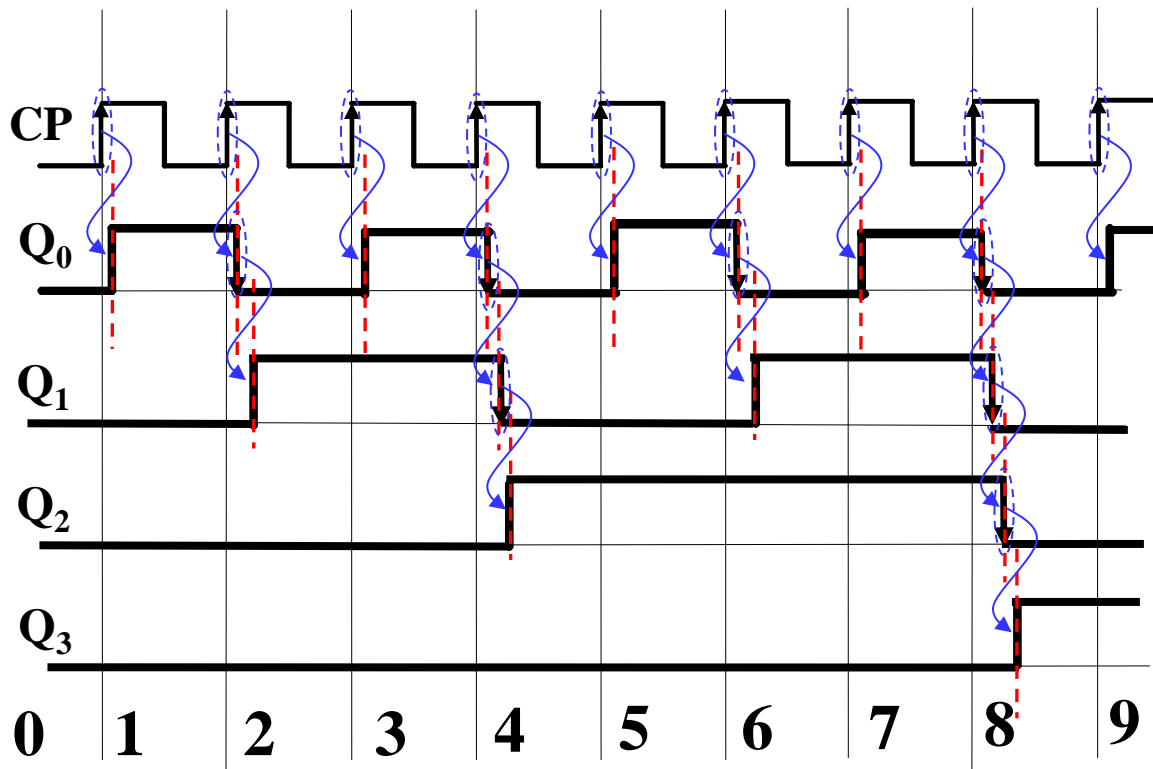
- How does it work?
 - When there is a positive edge on the clock input of A, A complements
 - The clock input for flip-flop B is the complemented output of flip-flop A
 - When flip A changes from 1 to 0, there is a positive edge on the clock input of B causing B to complement



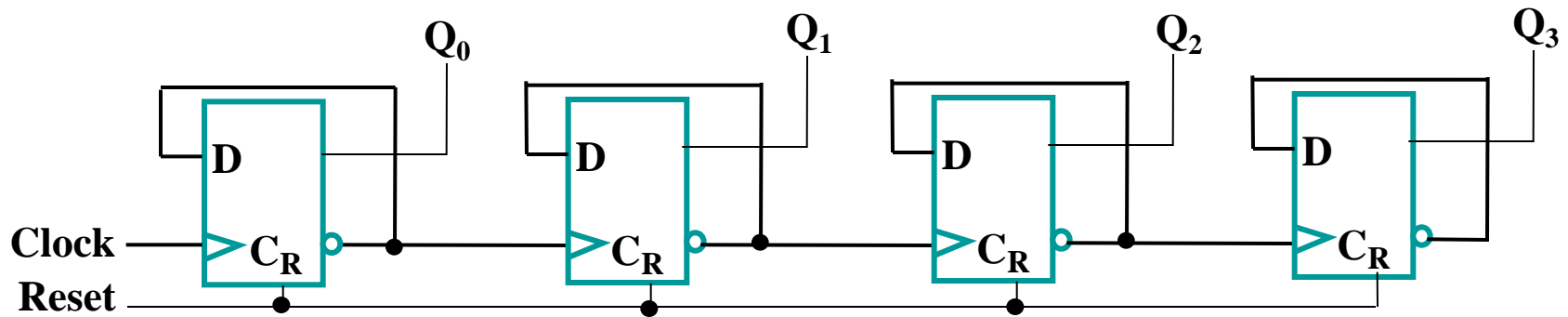
4-Bit Ripple Counter



4-Bit Ripple Counter

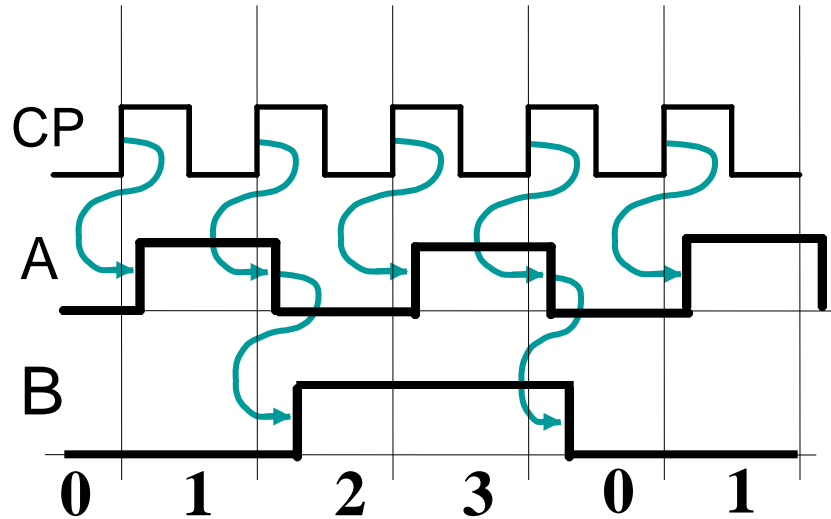


4-Bit Ripple Counter



Ripple Counter (continued)

- The arrows show the cause-effect relationship from the prior slide =>
- The corresponding sequence of states =>
 $(B,A) = (0,0), (0,1), (1,0), (1,1), (0,0), (0,1), \dots$
- Each additional bit, C, D, ...behaves like bit B, changing half as frequently as the bit before it.
- For 3 bits: $(C,B,A) = (0,0,0), (0,0,1), (0,1,0), (0,1,1), (1,0,0), (1,0,1), (1,1,0), (1,1,1), (0,0,0), \dots$



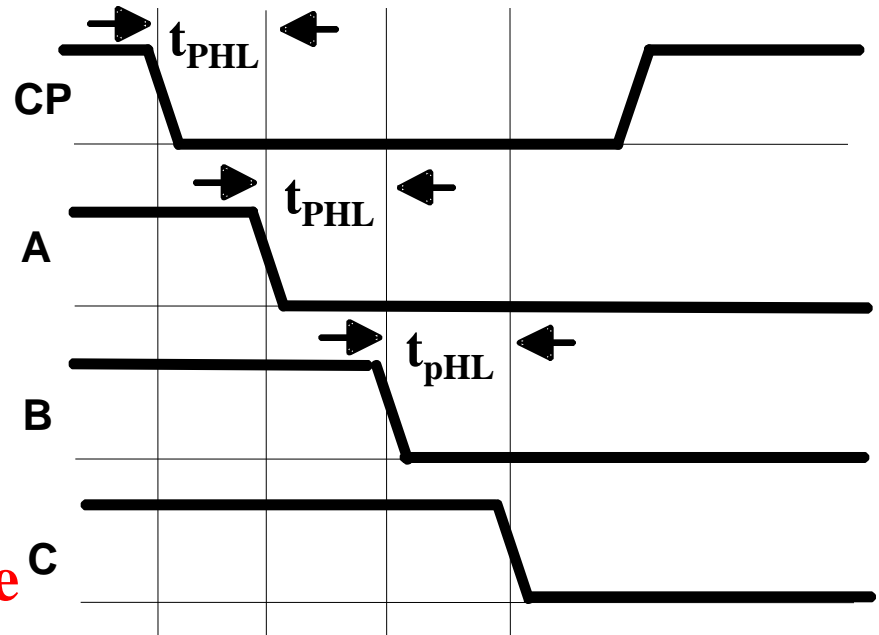
Ripple Counter (continued)

- These circuits are called *ripple counters* because each edge sensitive transition (positive in the example) causes a change in the next flip-flop's state.
- The changes “ripple” upward through the chain of flip-flops, i. e., each transition occurs after a clock-to-output delay from the stage before.

Ripple Counter (continued)

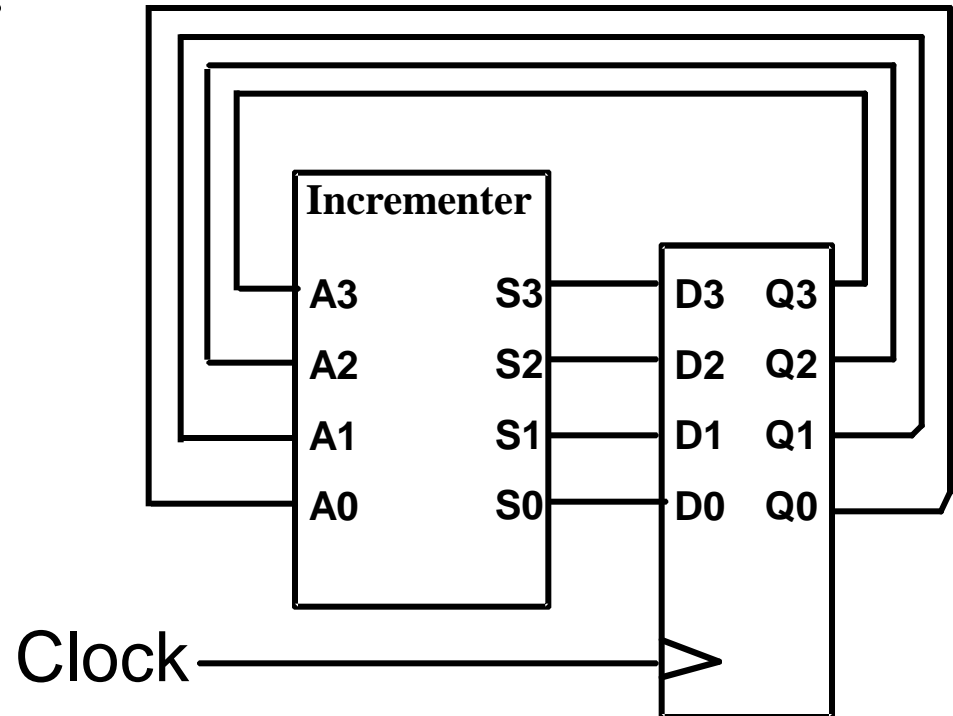
- Starting with $C = B = A = 1$, equivalent to $(C,B,A) = 7$ base 10, the next clock increments the count to $(C,B,A) = 0$ base 10. In fine timing detail:

- The clock to output delay t_{PHL} causes an increasing delay from clock edge for each stage transition.
- Thus, the count “ripples” from least to most significant bit.
- For n bits, total worst case delay is $n t_{PHL}$.**



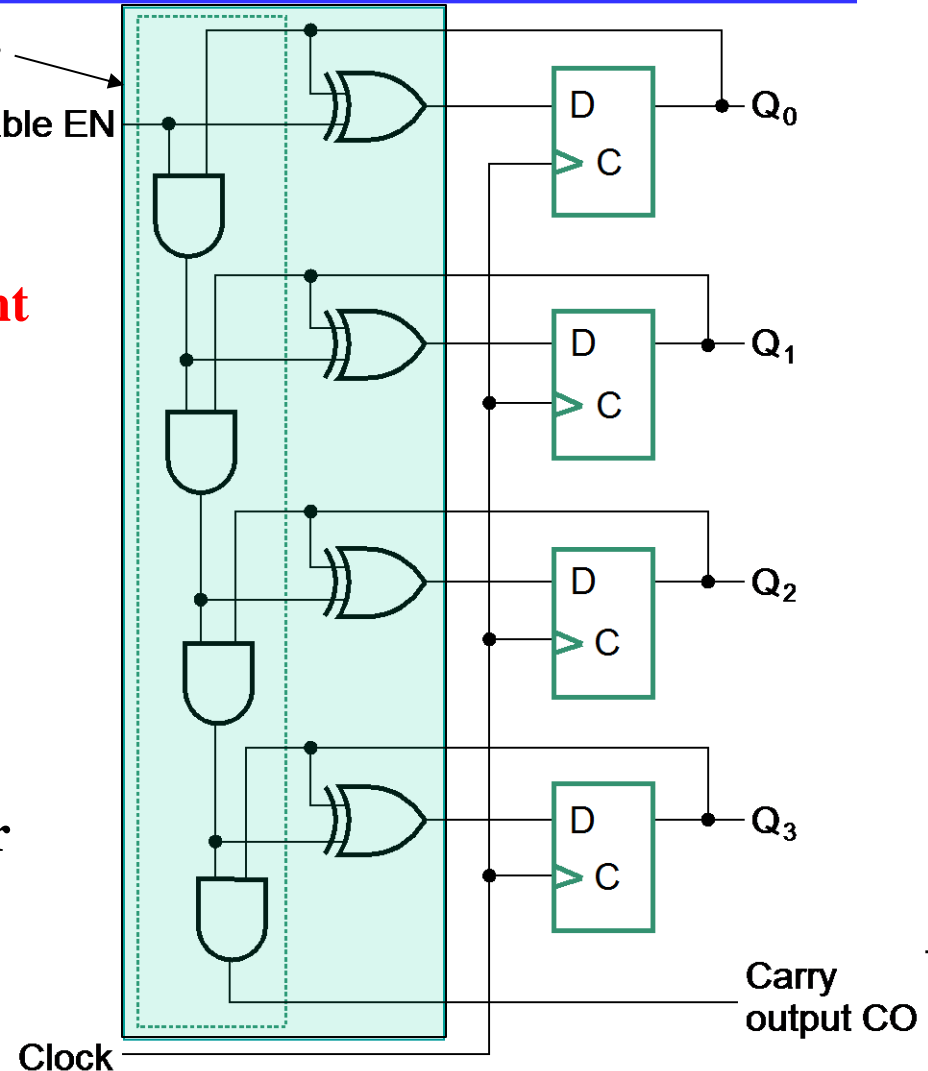
Synchronous Counters

- **To eliminate the "ripple" effects**, use a common clock for each flip-flop and a combinational circuit to generate the next state.
- For an up-counter, use an **incrementer**.



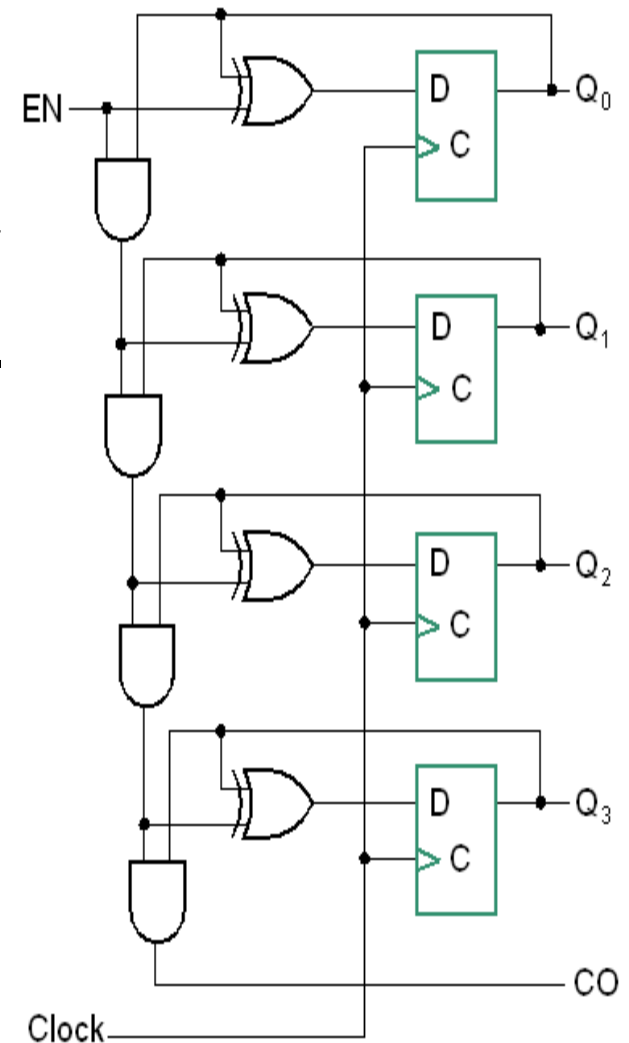
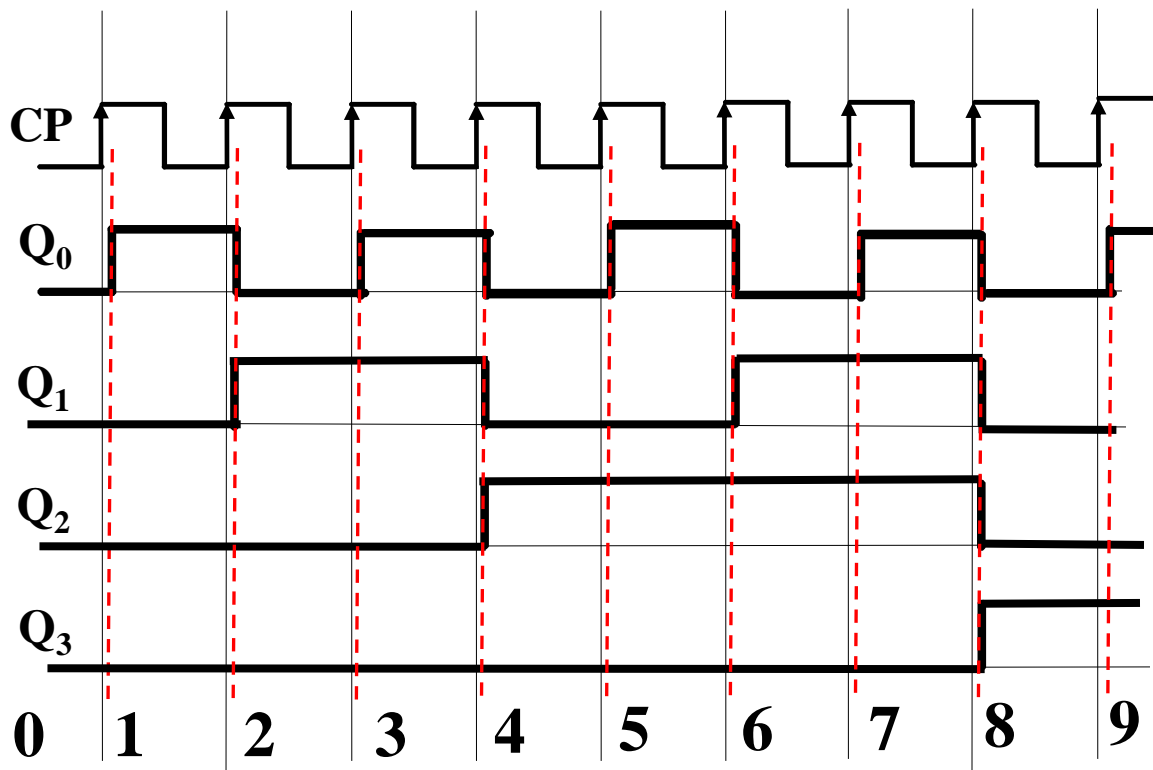
Synchronous Counters (continued)

- **Internal details => Incrementer**
- **Internal Logic**
 - **XOR complements each bit**
 - **AND chain causes complement of a bit if all bits toward LSB from it equal 1**
- **Count Enable**
 - Forces all outputs of AND chain to 0 to “hold” the state
- **Carry Out**
 - Added as part of incrementer
 - Connect to Count Enable of additional 4-bit counters to form larger counters



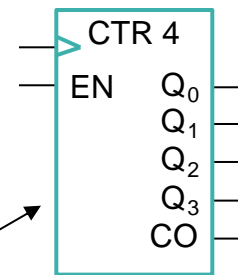
(a) Logic Diagram-Serial Gating

Synchronous Counters

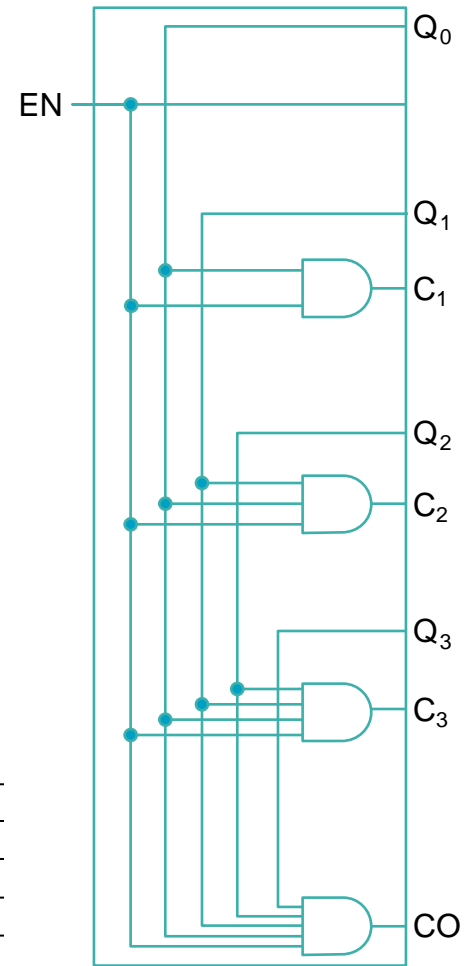


Synchronous Counters (continued)

- **Carry chain**
 - series of AND gates through which the carry “ripples”
 - Yields long path delays
 - Called *serial gating*
- **Replace AND carry chain with ANDs => in parallel**
 - Reduces path delays
 - Called *parallel gating*
 - Like carry lookahead
 - Lookahead can be used on COs and ENs to prevent long paths in large counters
- **Symbol for Synchronous Counter**



Symbol



Logic Diagram-Parallel Gating

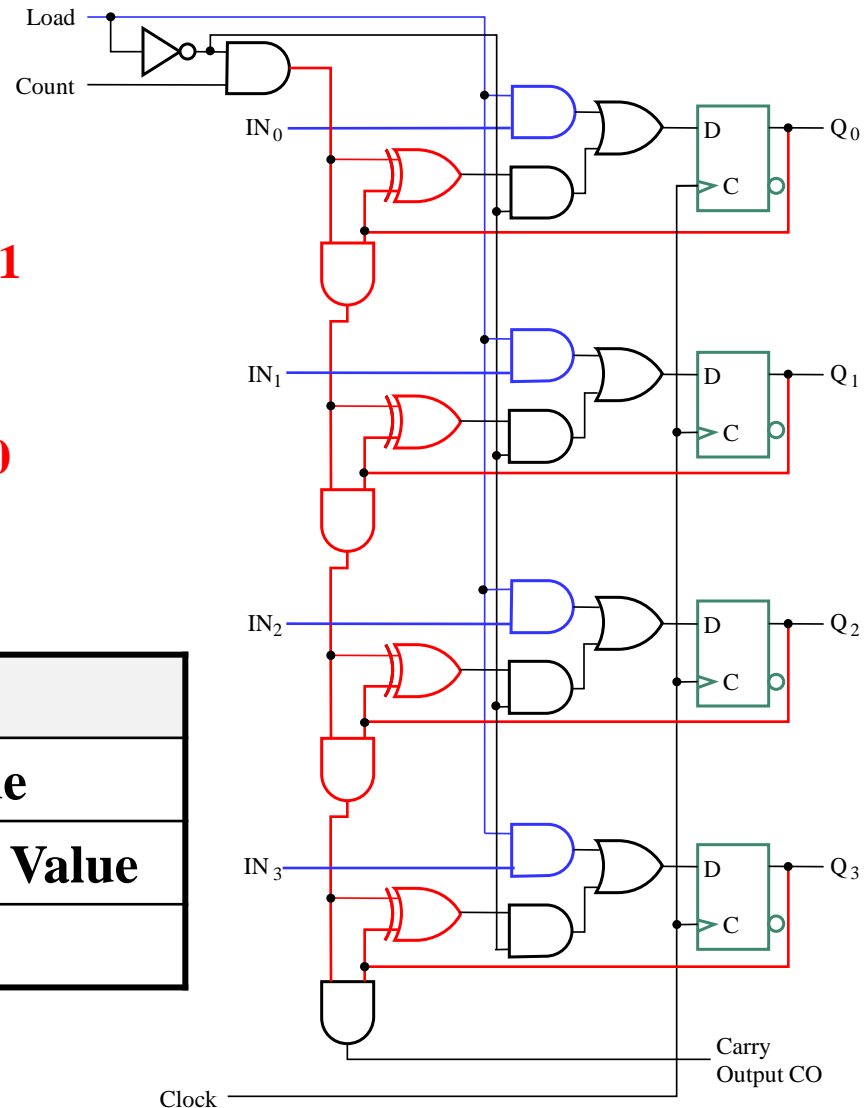
Other Counters

- See text for:
 - ***Down Counter*** - counts downward instead of upward
 - ***Up-Down Counter*** - counts up or down depending on value a control input such as Up/Down
 - ***Parallel Load Counter*** - Has parallel load of values available depending on control input such as Load
- ***Divide-by- n (Modulo n) Counter***
 - Count is remainder of division by n ; n may not be a power of 2 or
 - Count is arbitrary sequence of n states specifically designed state-by-state
 - Includes modulo 10 which is the *BCD counter*

Counter with Parallel Load

- Add path for input data
 - enabled for Load = 1
- Add logic to:
 - disable count logic for Load = 1
 - disable feedback from outputs for Load = 1
 - enable count logic for Load = 0 and Count = 1
- The resulting function table:

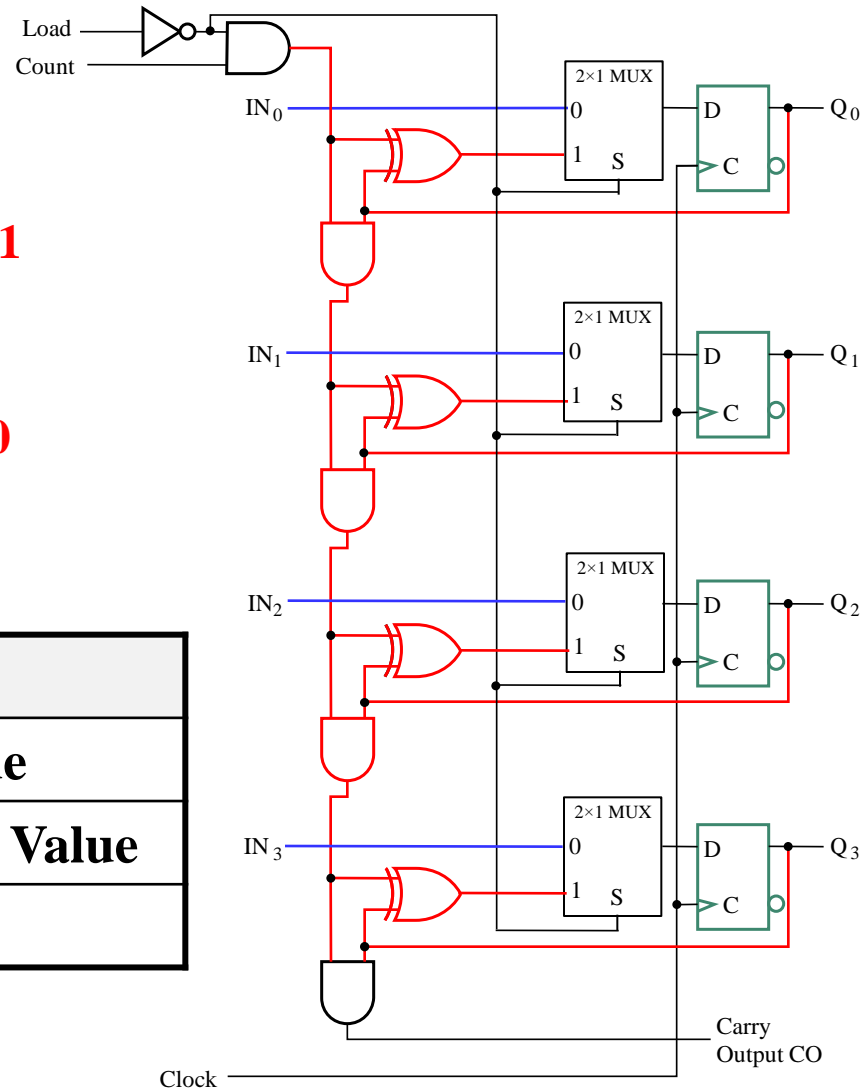
Load	Count	Action
0	0	Hold Stored Value
0	1	Count Up Stored Value
1	X	Load D



Counter with Parallel Load

- Add path for input data
 - enabled for Load = 1
- Add logic to:
 - disable count logic for Load = 1
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- The resulting function table:

Load	Count	Action
0	0	Hold Stored Value
0	1	Count Up Stored Value
1	X	Load D



Design Example: Synchronous BCD

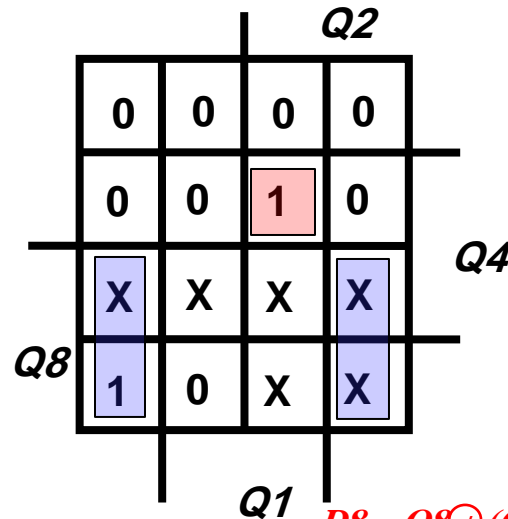
- Use the sequential logic model to design a synchronous BCD counter with D flip-flops
- State Table =>
- Input combinations 1010 through 1111 are don't cares

Current State				Next State			
Q8	Q4	Q2	Q1	Q8	Q4	Q2	Q1
0	0	0	0	0	0	0	1
0	0	0	1	0	0	1	0
0	0	1	0	0	0	1	1
0	0	1	1	0	1	0	0
0	1	0	0	0	1	0	1
0	1	0	1	0	1	1	0
0	1	1	0	0	1	1	1
0	1	1	1	1	0	0	0
1	0	0	0	1	0	0	1
1	0	0	1	0	0	0	0

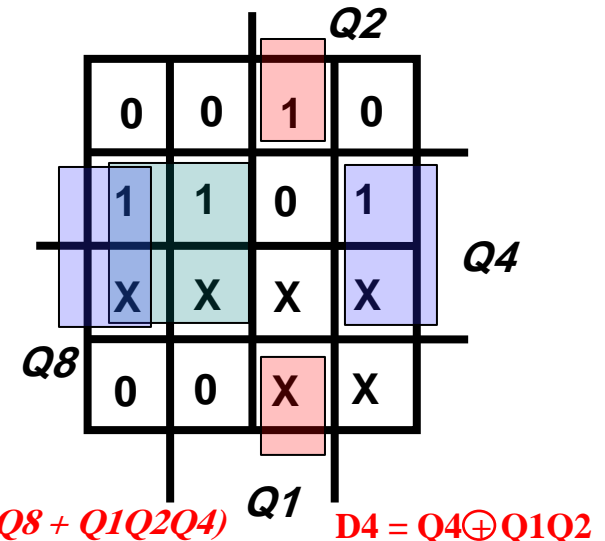
Synchronous BCD (continued)

Current State Q8 Q4 Q2 Q1	Next State Q8 Q4 Q2 Q1
0 0 0 0	0 0 0 1
0 0 0 1	0 0 1 0
0 0 1 0	0 0 1 1
0 0 1 1	0 1 0 0
0 1 0 0	0 1 0 1
0 1 0 1	0 1 1 0
0 1 1 0	0 1 1 1
0 1 1 1	1 0 0 0
1 0 0 0	1 0 0 1
1 0 0 1	0 0 0 0

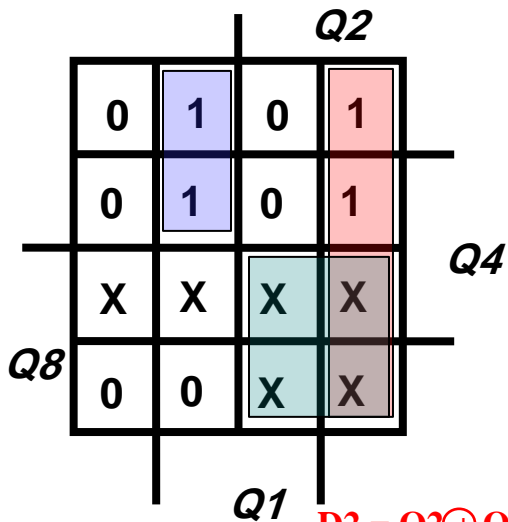
Present State Q8 Q4 Q2 Q1	Next State Q8 Q4 Q2 Q1
1 0 1 0	1 0 1 1
1 0 1 1	0 1 1 0
1 1 0 0	1 1 0 1
1 1 0 1	0 1 0 0
1 1 1 0	1 1 1 1
1 1 1 1	0 0 1 0



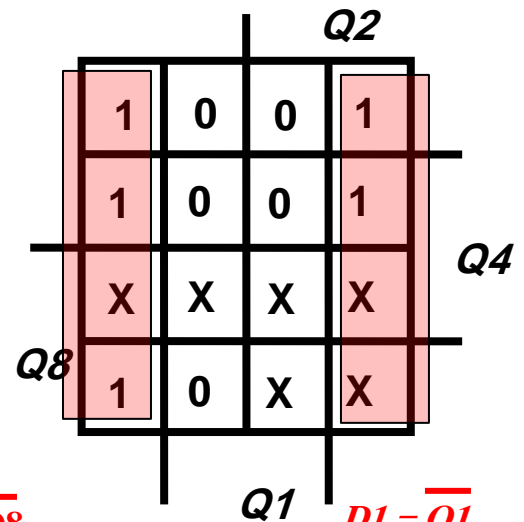
$$D8 = Q8 \oplus (Q1Q8 + Q1Q2Q4)$$



$$D4 = Q4 \oplus Q1Q2$$



$$D2 = Q2 \oplus Q1\overline{Q8}$$



$$D1 = \overline{Q1}$$

Synchronous BCD (continued)

- Use K-Maps to two-level optimize the next state equations and manipulate into forms containing XOR gates:

$$D1 = \overline{Q_1}$$

$$D2 = Q_2 \oplus Q_1 \overline{Q_8}$$

$$D4 = Q_4 \oplus Q_1 Q_2$$

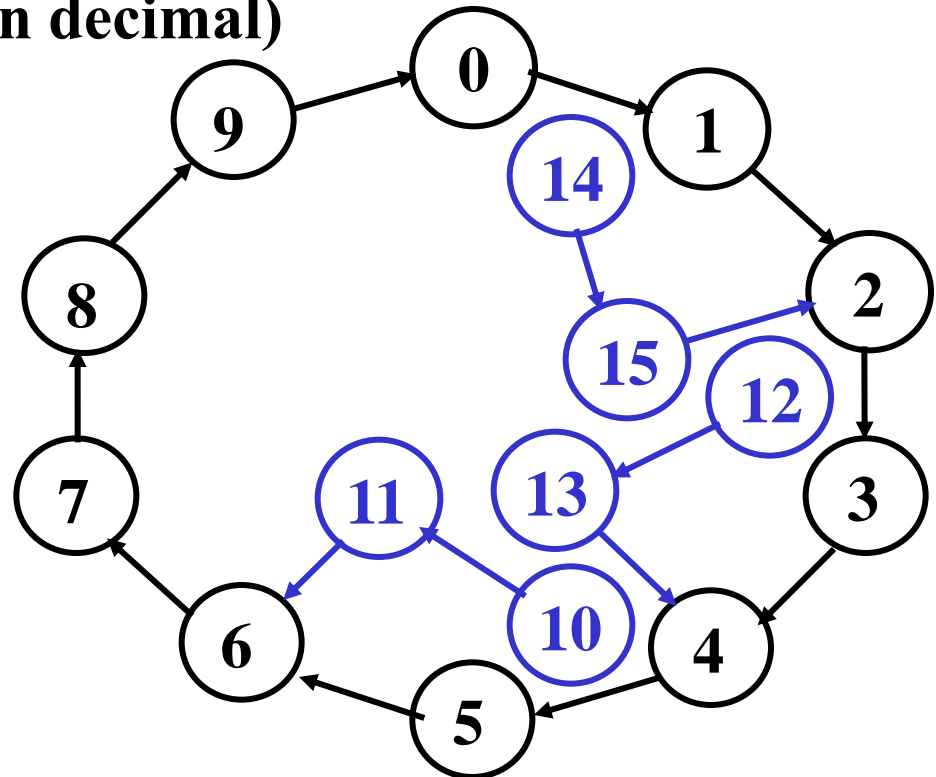
$$D8 = Q_8 \oplus (Q_1 Q_8 + Q_1 Q_2 Q_4)$$

- The logic diagram can be draw from these equations
 - An asynchronous or synchronous reset should be added
- What happens if the counter is perturbed by a **power disturbance** or other interference and it enters a state **other than 0000 through 1001**?

Synchronous BCD (continued)

- Find the actual values of the six next states for the don't care combinations from the equations
- Find the overall state diagram to assess behavior for the don't care states (states in decimal)

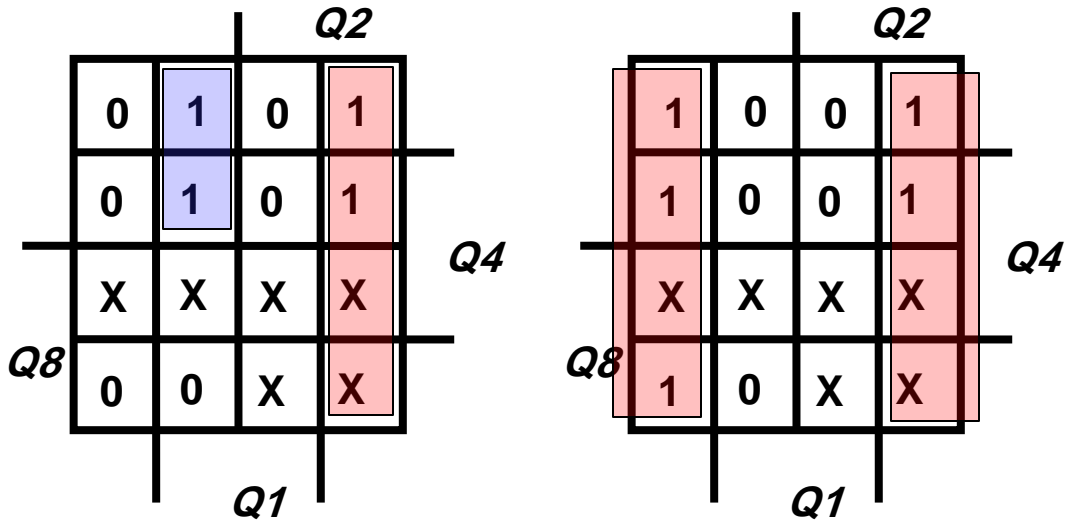
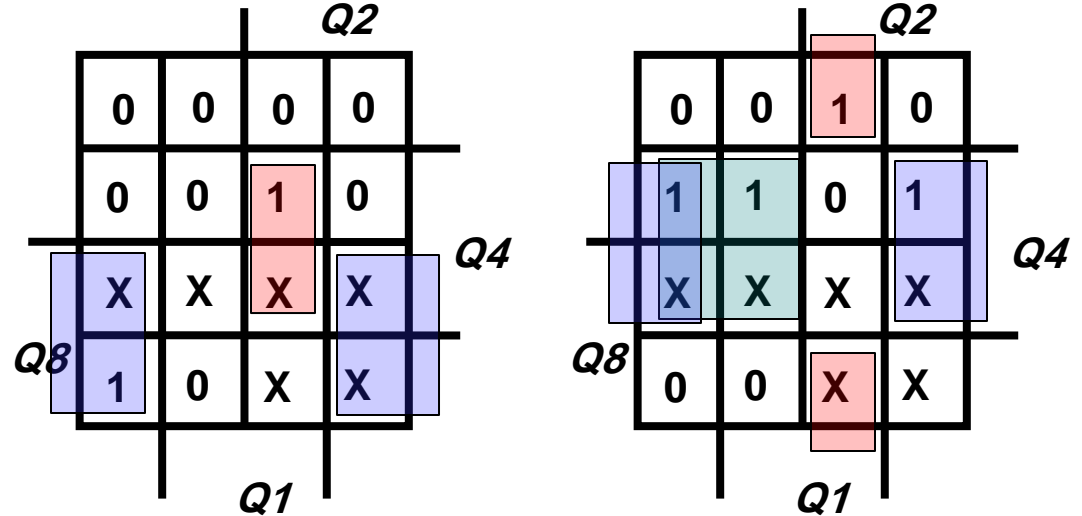
Present State				Next State			
Q8	Q4	Q2	Q1	Q8	Q4	Q2	Q1
1	0	1	0	1	0	1	1
1	0	1	1	0	1	1	0
1	1	0	0	1	1	0	1
1	1	0	1	0	1	0	0
1	1	1	0	1	1	1	1
1	1	1	1	0	0	1	0



Synchronous BCD (Another Solution)

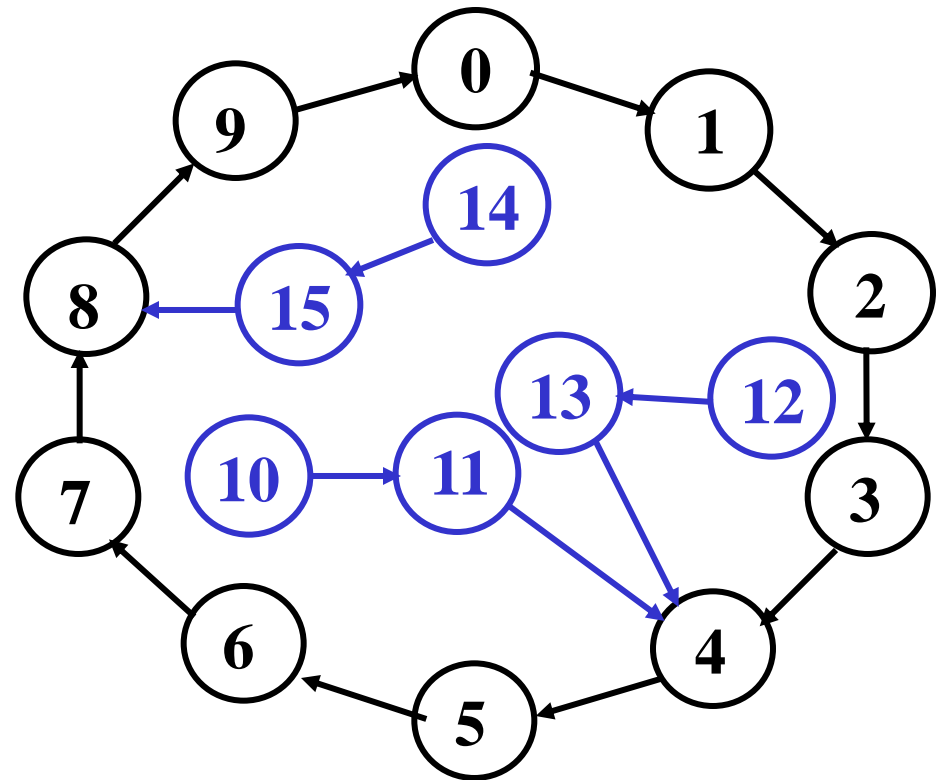
Current State Q8 Q4 Q2 Q1	Next State Q8 Q4 Q2 Q1
0 0 0 0	0 0 0 1
0 0 0 1	0 0 1 0
0 0 1 0	0 0 1 1
0 0 1 1	0 1 0 0
0 1 0 0	0 1 0 1
0 1 0 1	0 1 1 0
0 1 1 0	0 1 1 1
0 1 1 1	1 0 0 0
1 0 0 0	1 0 0 1
1 0 0 1	0 0 0 0

Present State	Next State
Q8 Q4 Q2 Q1	Q8 Q4 Q2 Q1
1 0 1 0	1 0 1 1
1 0 1 1	0 1 0 0
1 1 0 0	1 1 0 1
1 1 0 1	0 1 0 0
1 1 1 0	1 1 1 1
1 1 1 1	1 0 0 0



Synchronous BCD (Another Solution)

Present State				Next State			
Q8	Q4	Q2	Q1	Q8	Q4	Q2	Q1
1	0	1	0	1	0	1	1
1	0	1	1	0	1	0	0
1	1	0	0	1	1	0	1
1	1	0	1	0	1	0	0
1	1	1	0	1	1	1	1
1	1	1	1	1	0	0	0



Synchronous BCD (continued)

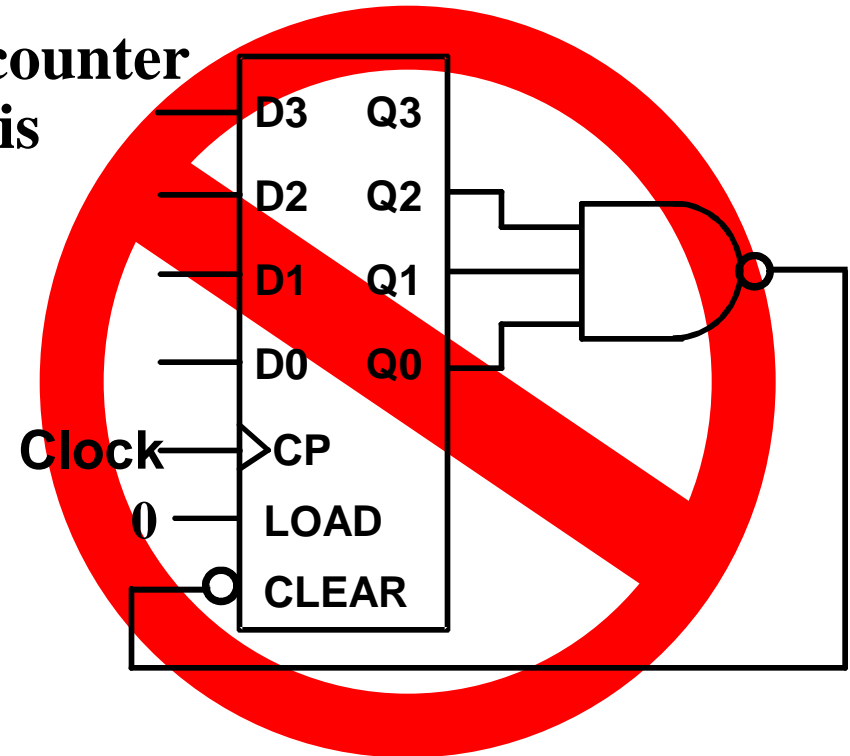
- For the BCD counter design, if an invalid state is entered, **return to a valid state occurs within two clock cycles**
- Is this adequate? If not:
 - Is a signal needed that indicates that an invalid state has been entered? What is the equation for such a signal?
 - Does the design need to be modified to return from an invalid state to a valid state in one clock cycle?
 - Does the design need to be modified to return from a invalid state to a specific state (such as 0)?
- The action to be taken depends on:
 - the application of the circuit
 - design group policy

Counting Modulo N

- The following techniques use an n -bit binary counter with asynchronous or synchronous clear and/or parallel load:
 - Detect a *terminal count* of N in a Modulo-N count sequence to asynchronously Clear the count to 0 or asynchronously Load in value 0 (These lead to counts which are present for only a very short time and can fail to work for some timing conditions!)
 - Detect a terminal count of N - 1 in a Modulo-N count sequence to Clear the count synchronously to 0
 - Detect a terminal count of N - 1 in a Modulo-N count sequence to synchronously Load in value 0
 - Detect a terminal count and use Load to preset a count of the terminal count value minus (N - 1)
- Alternatively, custom design a modulo N counter as done for BCD

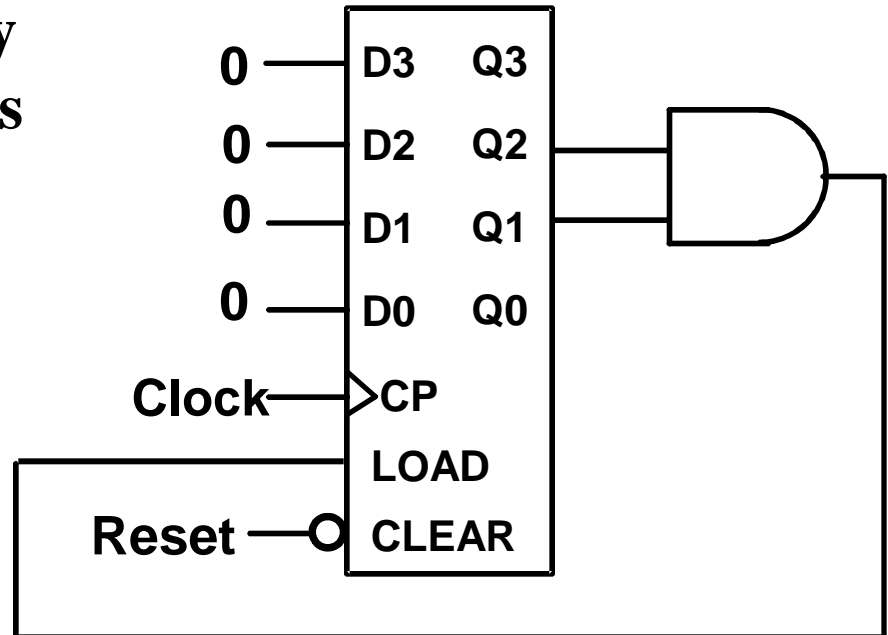
Counting Modulo 7: Detect 7 and Asynchronously Clear

- A synchronous 4-bit binary counter with an asynchronous Clear is used to make a Modulo 7 counter.
- Use the Clear feature to detect the count 7 and clear the count to 0. This gives a count of 0, 1, 2, 3, 4, 5, 6, **7(short)0**, 1, 2, 3, 4, 5, 6, **7(short)0**, etc.
- **DON'T DO THIS!** Existence of state 7 may not be long enough to reliably reset all flip-flops to 0. Referred to as a “suicide” counter! (Count “7” is “killed,” but the designer’s job may be dead as well!)

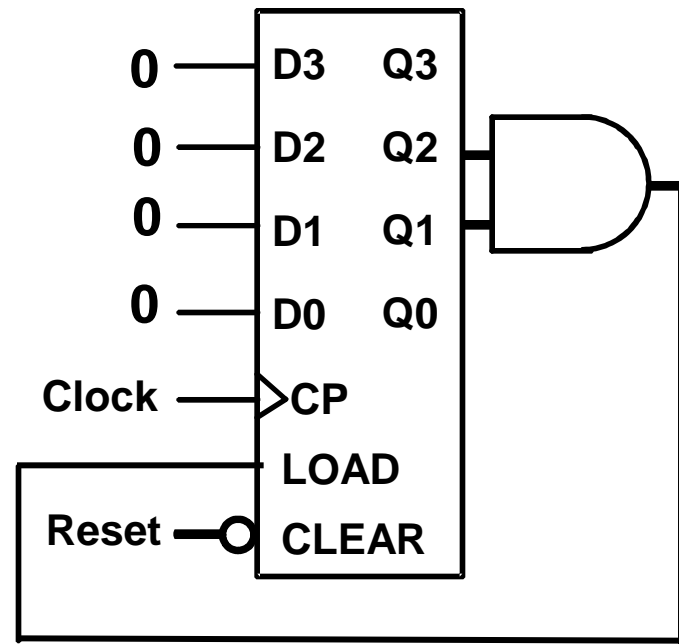
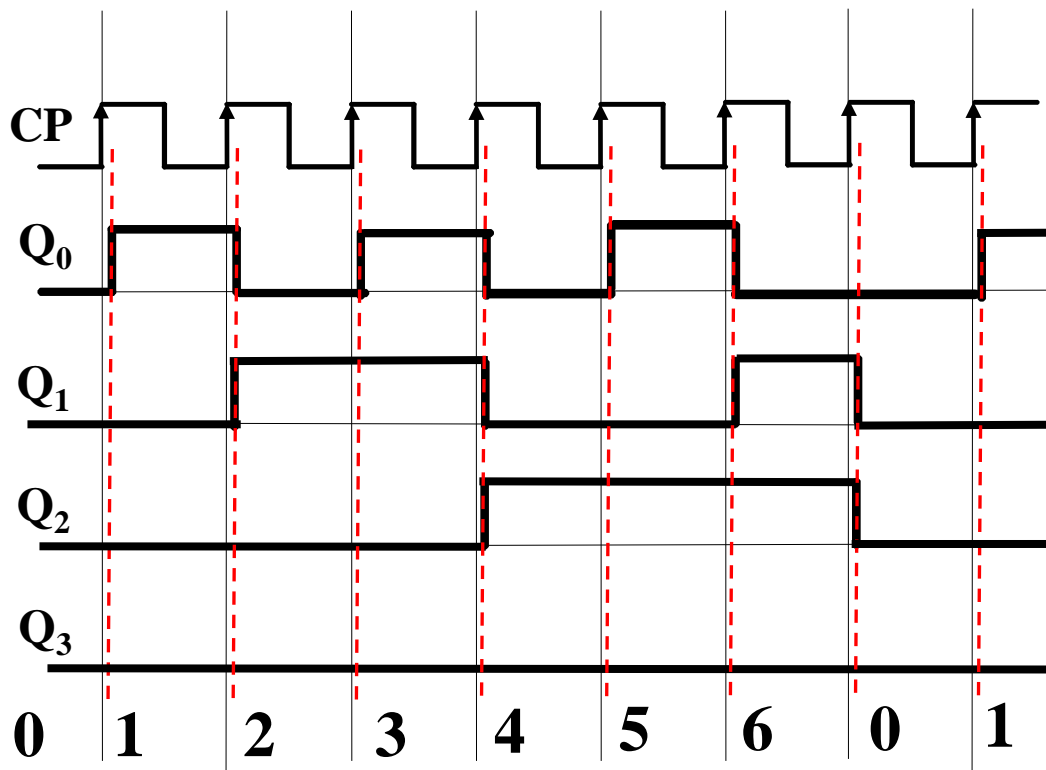


Counting Modulo 7: Synchronously Load on Terminal Count of 6

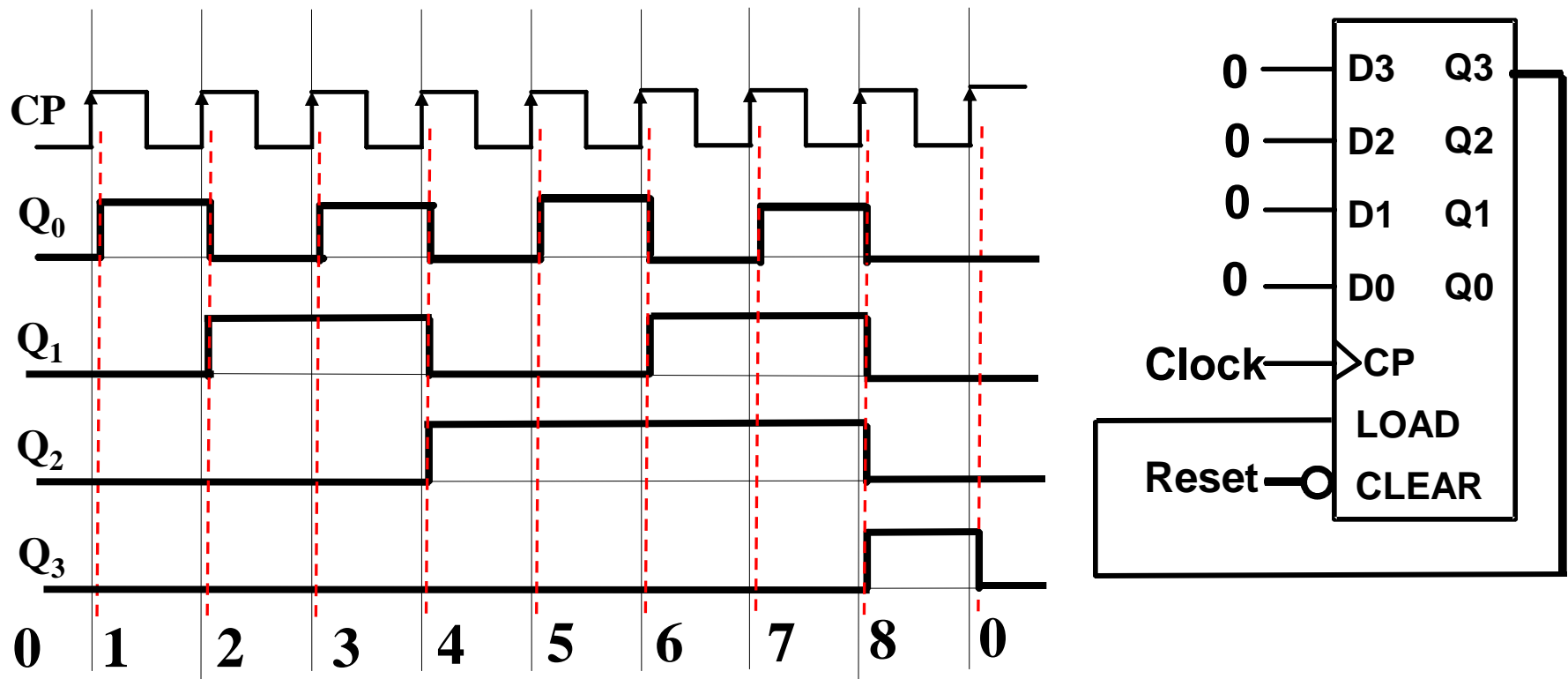
- A synchronous 4-bit binary counter with a synchronous load and an asynchronous clear is used to make a **Modulo 7 counter**
- Use the Load feature to detect the count "6" and load in "zero". This gives a count of **0, 1, 2, 3, 4, 5, 6, 0, 1, 2, 3, 4, 5, 6, 0, ...**
- Using don't cares for states above 0110, detection of 6 can be done with **Load = Q4 Q2**



Counting Modulo 7: Synchronously Load on Terminal Count of 6

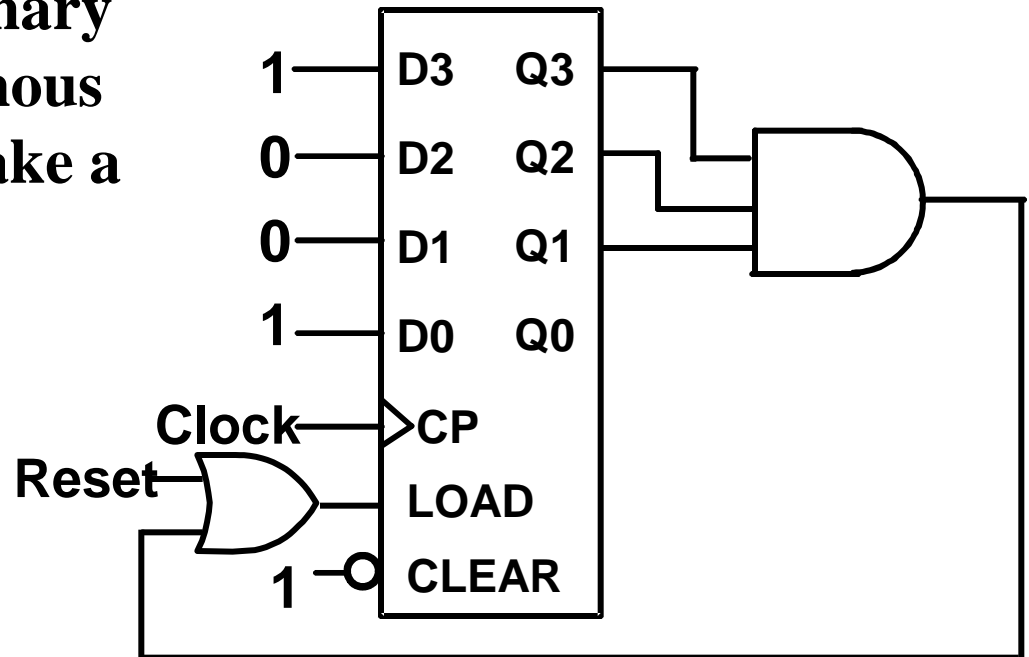


Counting Modulo 9: Synchronously Load on Terminal Count of 8



Counting Modulo 6: Synchronously Preset 9 on Reset and Load 9 on Terminal Count 14

- A synchronous, 4-bit binary counter with a synchronous Load is to be used to make a Modulo 6 counter.
- Use the Load feature to preset the count to 9 on Reset and detection of count 14.



- This gives a count of 9, 10, 11, 12, 13, 14, 9, 10, 11, 12, 13, 14, 9, ...
- If the terminal count is 15 detection is usually built in as Carry Out (CO)

Register Cell Design

- Assume that a register consists of identical cells
- Then register design can be approached as follows:
 - Design representative cell for the register
 - Connect copies of the cell together to form the register
 - Applying appropriate “boundary conditions” to cells that need to be different and contract if appropriate
- Register cell design is the first step of the above process

Register Cell Specifications

- A register
- Data inputs to the register
- Control input combinations to the register
 - **Example 1: Not encoded**
 - Control inputs: **Load, Shift, Add**
 - At most, one of Load, Shift, Add is 1 for any clock cycle
(0,0,0), (1,0,0), (0,1,0), (0,0,1)
 - **Example 2: Encoded**
 - Control inputs: **S1, S0**
 - All possible binary combinations on S1, S0
(0,0), (0,1), (1,0), (1,1)

Register Cell Specifications

- A set of register functions (typically specified as register transfers)

- **Example:**

Load: $A \leftarrow B$

Shift: $A \leftarrow \text{sr } B$

Add: $A \leftarrow A + B$

- A hold state specification

- **Example:**

- Control inputs: Load, Shift, Add
- **If all control inputs are 0, hold the current register state**

Example 1: Register Cell Design

- **Register A (m-bits) Specification:**
 - **Data input: B**
 - **Control inputs (CX, CY)**
 - **Control input combinations $(0,0), (0,1) (1,0)$**
 - **Register transfers:**
 - **$CX: A \leftarrow B \vee A$**
 - **$CY : A \leftarrow B \oplus A$**
 - **Hold state: $(0,0)$**

Example 1: Register Cell Design (continued)

- Load Control

$$\text{Load} = \text{CX} + \text{CY}$$

- All control combinations appear as if encoded (0,0), (0,1), (1,0).

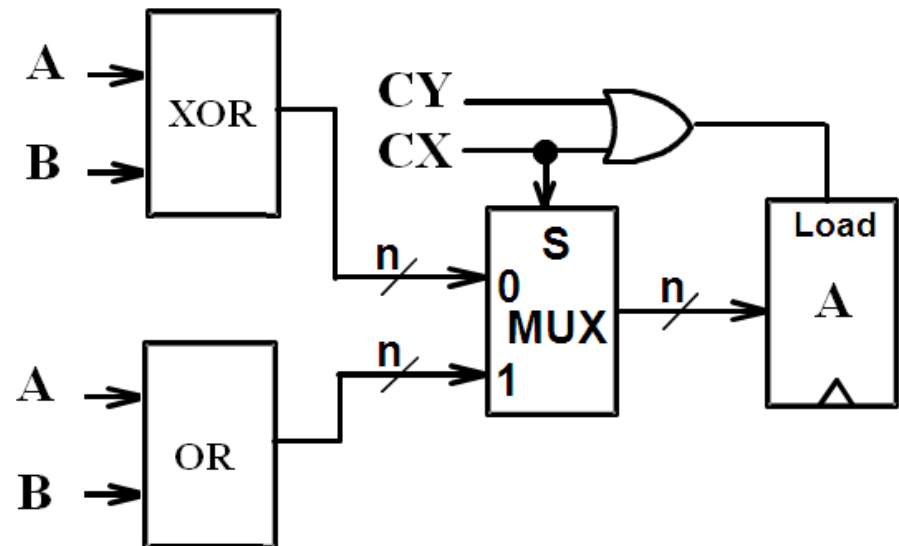
$$S = CX$$

$$A_i \leftarrow B_i \oplus A_i$$

$$A_i \leftarrow B_i \vee A_i$$

$$CY = 1$$

$$CX = 1$$



Sequential Circuit Design Approach

- **Find a state diagram or state table**
 - **Note that there are only two states with the state assignment equal to the register cell output value**
- **Use the design procedure in Chapter 5 to complete the cell design**
- **For optimization:**
 - **Use K-maps for up to 4 to 6 variables**
 - **Otherwise, use computer-aided or manual optimization**

Example 1 Again

■ State Table:

	Hold	$A_i \vee B_i$		$A_i \oplus B_i$	
	$CX = 0$ $CY = 0$	$CX = 1$ $B_i = 0$	$CX = 1$ $B_i = 1$	$CY = 1$ $B_i = 0$	$CY = 1$ $B_i = 1$
A_i					
0	0	0	1	0	1
1	1	1	1	1	0

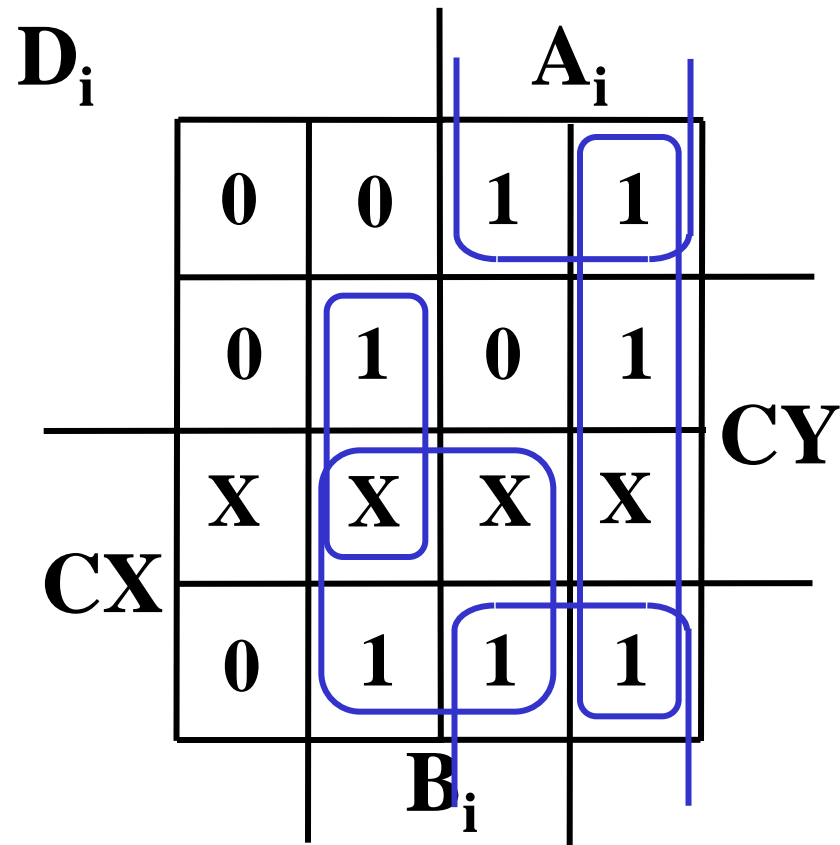
- Four variables give a total of 16 state table entries
- By using:
 - Combinations of variable names and values
 - Don't care conditions (for $CX = CY = 1$)

only 8 entries are required to represent the 16 entries

Example 1 Again (continued)

- K-map - Use variable ordering CX, CY, A_i B_i and assume a D flip-flop**

CX	CY	A _i	B _i	D _i	Description
0	0	0	0	0	HOLD
0	0	0	1	0	
0	0	1	0	1	
0	0	1	1	1	
0	1	0	0	0	CY: $A \leftarrow B \oplus A$
0	1	0	1	1	
0	1	1	0	1	
0	1	1	1	0	
1	0	0	0	0	CX: $A \leftarrow B \vee A$
1	0	0	1	1	
1	0	1	0	1	
1	0	1	1	1	
1	1	0	0	X	No Action
1	1	0	1	X	
1	1	1	0	X	
1	1	1	1	X	



Example 1 Again (continued)

- The resulting SOP equation:

$$D_i = CX B_i + CY \bar{A}_i B_i + A_i \bar{B}_i + \bar{C}\bar{Y} A_i$$

- Using factoring and DeMorgan's law:

$$D_i = CX B_i + \bar{A}_i (CY B_i) + A_i (\overline{CY B_i})$$

$$D_i = CX B_i + A_i \oplus (CY B_i)$$

The gate input cost per cell = $2 + 8 + 2 + 2 = 14$

- The gate input cost per cell for the previous version is:

Per cell: 19

Shared decoder logic: 8

- Cost gain by sequential design > 5 per cell
- Also, no Enable on the flip-flop makes it cost less

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