

# Adaptive Filter

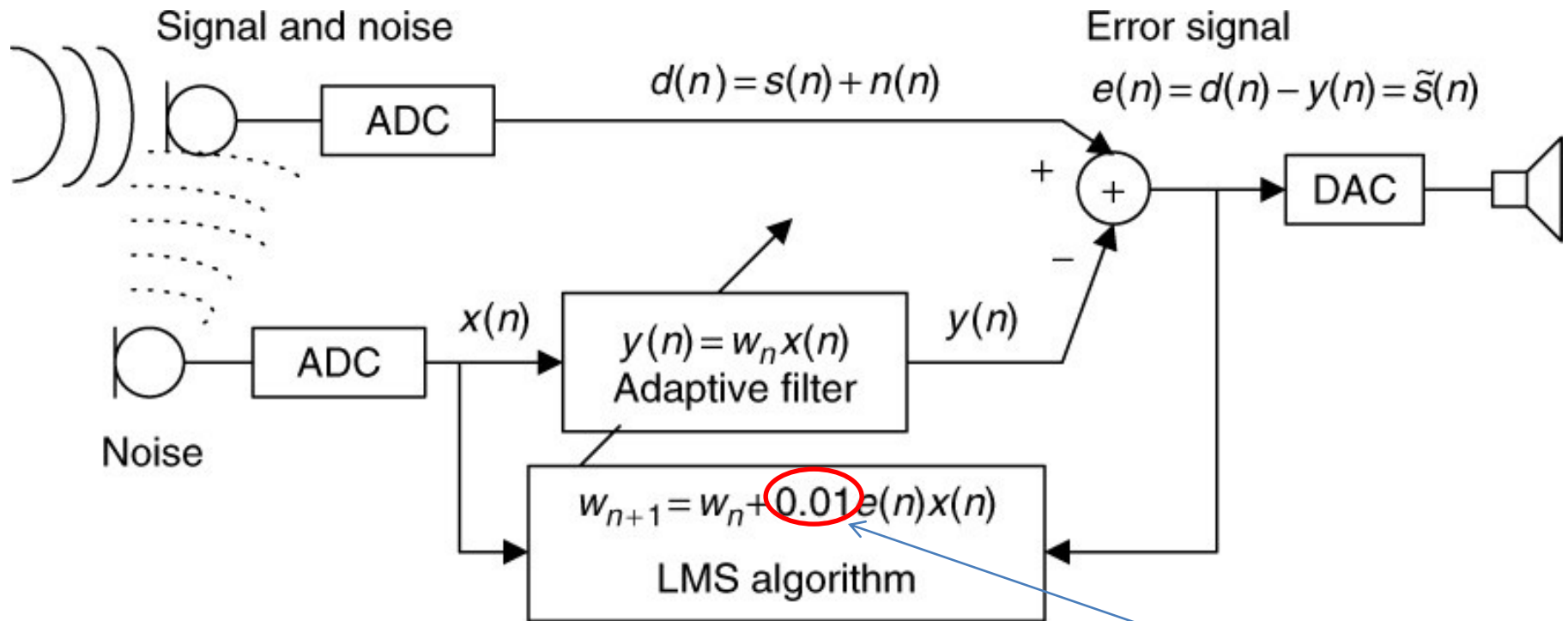
A digital filter that automatically **adjusts its coefficients** to adapt input signal **via an adaptive algorithm**.

## Applications:

- **Signal enhancement**
- **Active noise control**
- **Noise cancellation**
- **Telephone echo cancellation**

Text: Digital Signal Processing  
by Li Tan, Chapter 10

# Simplest Noise Canceller



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$n(n)$  is a linear filtered (delayed) version of  $x(n)$ .

Controls speed of convergence

# Simplest Noise Canceller - contd.

$$y(n) = w_n x(n)$$

$$e(n) = d(n) - y(n)$$

Initial coefficient  $w_0 = 0.3$

$$w_{n+1} = w_n + 0.01e(n)x(n).$$

$n$	$d(n)$	$x(n)$
0	-0.2947	-0.5893
1	1.0017	0.5893

← Measured

$$n = 0, \quad y(0) = w_0 x(0) = 0.3 \times (-0.5893) = -0.1768$$

$$e(0) = d(0) - y(0) = -0.2947 - (-0.1768) = -0.1179 = \tilde{s}(0)$$

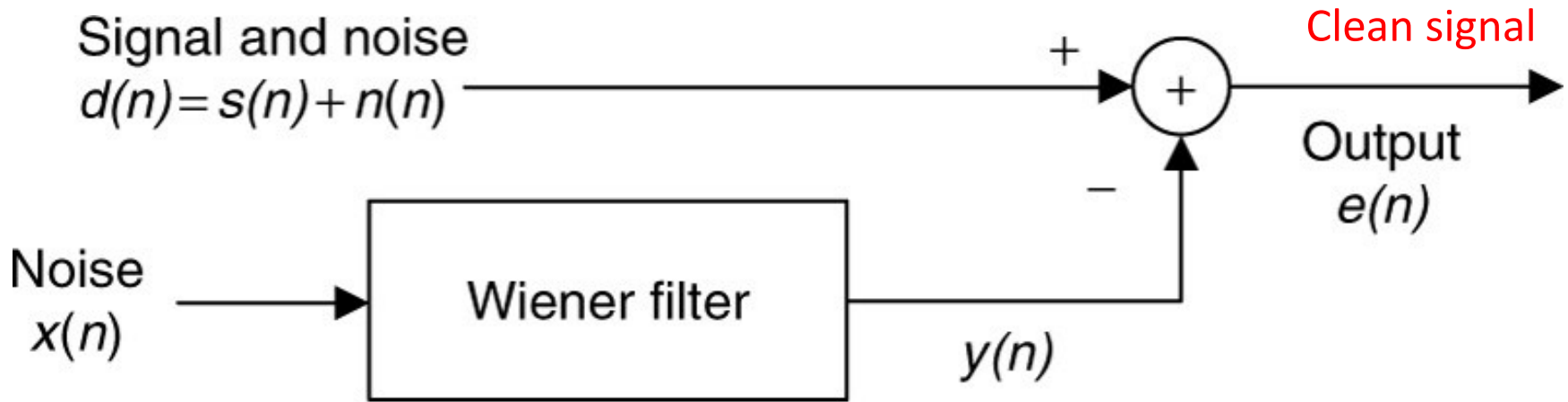
$$w_1 = w_0 + 0.01e(0)x(0) = 0.3 + 0.01 \times (-0.1179) \times (-0.5893) = 0.3007$$

$$n = 1, \quad y(1) = w_1 x(1) = 0.3007 \times 0.5893 = 0.1772$$

$$e(1) = d(1) - y(1) = 1.0017 - 0.1772 = 0.8245 = \tilde{s}(1)$$

$$w_2 = w_1 + 0.01e(1)x(1) = 0.3007 + 0.01 \times 0.8245 \times 0.5893 = 0.3056$$

# Wiener Filter & LMS Algorithm



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Consider, single weight case,  $y(n) = wx(n)$ .

Error signal,  $e(n) = d(n) - wx(n)$

Now we have to solve for the best weight  $w^*$

$$e^2(n) = (d(n) - wx(n))^2 = d^2(n) - 2d(n)wx(n) + w^2x^2(n)$$

# LMS Algorithm

Taking Expectation of squared error signal

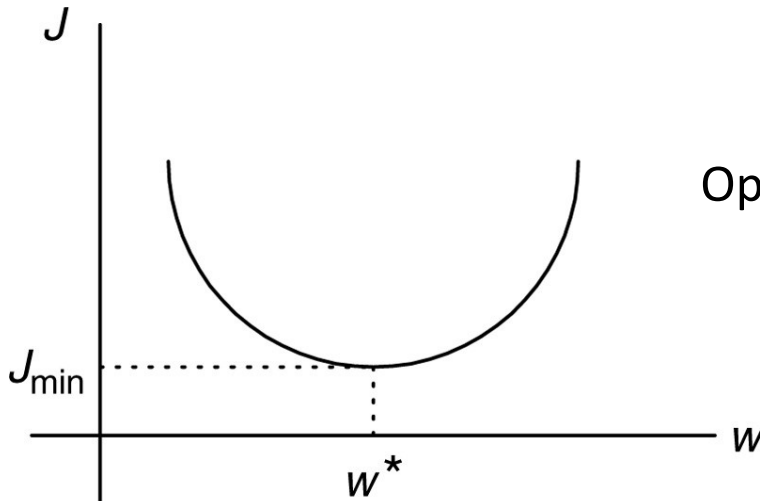
$$E(e^2(n)) = E(d^2(n)) - 2wE(d(n)x(n)) + w^2E(x^2(n))$$

$$J = E(e^2(n)) = \text{MSE (mean squared error)}$$

$$\sigma^2 = E(d^2(n)) = \text{power of corrupted signal}$$

$$P = E(d(n)x(n)) = \text{cross-correlation between } d(n) \text{ and } x(n)$$

$$R = E(x^2(n)) = \text{autocorrelation}$$



For large N,  $J = \sigma^2 - 2wP + w^2R$

Optimal  $w^*$  is found when minimum J is achieved

$$\frac{dJ}{dw} = -2P + 2wR = 0.$$

$$w^* = R^{-1}P$$

# LMS Algorithm - Example

Given MSE function for the Wiener filter:

$$\frac{dJ}{dw} = -20 + 10 \times 2w = 0$$

Solving for optimal, we get  $w^* = 1$

Finally we get

$$J_{\min} = J|_{w=w^*} = 40 - 20w + 10w^2|_{w=1} = 40 - 20 \times 1 + 10 \times 1^2 = 30$$

large N

$$w^* = R^{-1}P$$

$R^{-1}$ : Matrix inversion

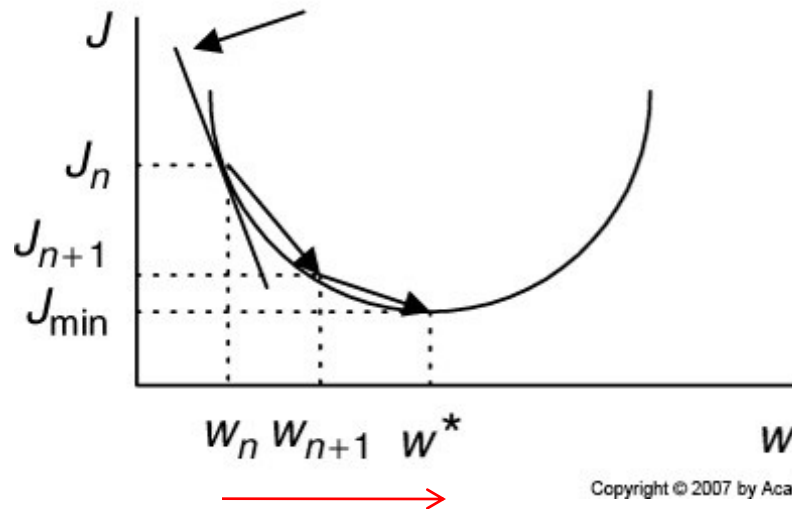
Makes real-time implementation difficult

# Steepest Decent Algorithm

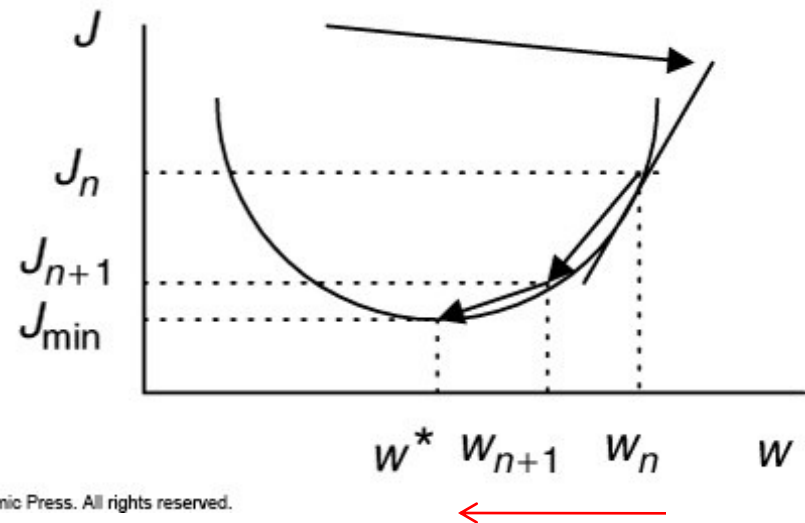
$$w_{n+1} = w_n - \mu \frac{dJ}{dw}$$

$\mu$  = constant controlling the speed of convergence.

Case  $\frac{dJ}{dw} < 0$  and  $-\mu \frac{dJ}{dw} > 0$



Case  $\frac{dJ}{dw} > 0$  and  $-\mu \frac{dJ}{dw} < 0$



# Steepest Decent Algorithm: Example

Given:  $J = 40 - 20w + 10w^2$

$$\mu = 0.04 \quad w_0 = 0$$

Iteration three times

Find optimal solution for  $w^*$

Solution:

$$\frac{dJ}{dw} = -20 + 10 \times 2w_n$$

For  $n = 0$ ,

$$\mu \frac{dJ}{dw} = 0.04 \times (-20 + 10 \times 2w_0)|_{w_0=0} = -0.8$$

$$w_1 = w_0 - \mu \frac{dJ}{dw} = 0 - (-0.8) = 0.8$$

For  $n = 1$ ,

$$\mu \frac{dJ}{dw} = 0.04 \times (-20 + 10 \times 2w_1)|_{w_1=0.8} = -0.16$$

$$w_2 = w_1 - \mu \frac{dJ}{dw} = 0.8 - (-0.16) = 0.96$$

For  $n = 2$ ,

$$\mu \frac{dJ}{dw} = 0.04 \times (-20 + 10 \times 2w_2)|_{w_2=0.96} = -0.032$$

$$w_3 = w_2 - \mu \frac{dJ}{dw} = 0.96 - (-0.032) = 0.992.$$

$$J_{\min} \approx 40 - 20w + 10w^2|_{w=0.992} \\ = 30.0006$$

$$w^* \approx w_3 = 0.992$$



# Steepest Decent Algorithm - contd1.

To make it sample-based processing, we need to take out estimation.

$$J = e^2(n) = (d(n) - wx(n))^2$$
$$\frac{dJ}{dw} = 2(d(n) - wx(n)) \frac{d(d(n) - wx(n))}{dw} = -2e(n)x(n)$$

Updating weight  $w_{n+1} = w_n + 2\mu e(n)x(n)$

For multiple tap FIR filter:

$$y(n) = w_n(0)x(n) + w_n(1)x(n-1) + \dots + w_n(N-1)x(n-N+1)$$

for  $i = 0, \dots, N-1$

$$w_{n+1}(i) = w_n(i) + 2\mu e(n)x(n-i).$$

Choose convergence  
constant as

$$0 < \mu < \frac{1}{NP_x}$$

$P_x$  : maximum input power

# Steepest Decent Algorithm - contd2.

## Steps:

1. Initialize  $w(0), w(1), \dots, w(N - 1)$  to arbitrary values.

2. Read  $d(n), x(n)$ , and perform digital filtering:

$$y(n) = w(0)x(n) + w(1)x(n - 1) + \dots + w(N - 1)x(n - N + 1).$$

3. Compute the output error:

$$e(n) = d(n) - y(n).$$

4. Update each filter coefficient using the LMS algorithm:

for  $i = 0, \dots, N - 1$

$$w(i) = w(i) + 2\mu e(n)x(n - i).$$

# Noise Canceller Using Adaptive Filter-1

Perform adaptive filtering to obtain outputs  $e(n) = n = 0, 1, 2$

**Given:**

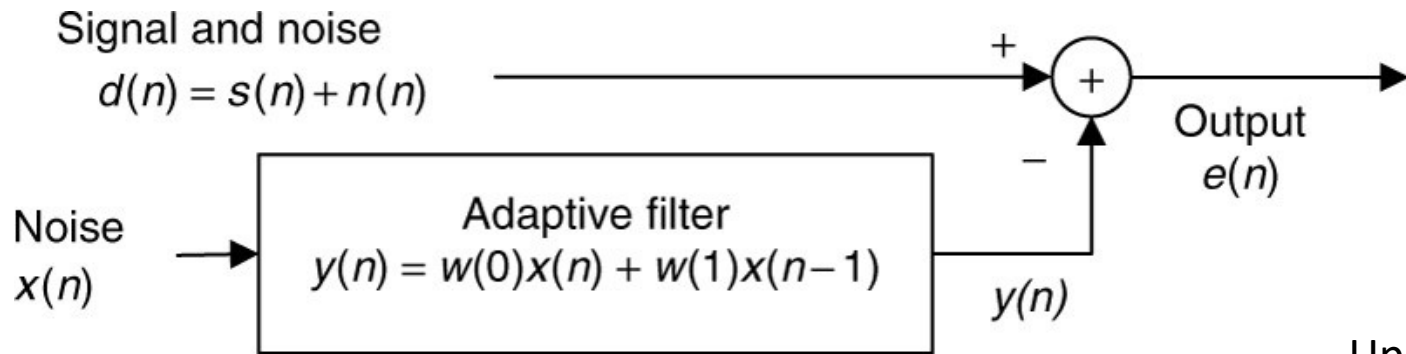
$$x(0) = 1, x(1) = 1, x(2) = -1,$$

$$d(0) = 2, d(1) = 1, d(2) = -2$$

Initial weights:

$$w(0) = w(1) = 0,$$

$$\mu = 0.1$$



Updating weights:

**Solution:**

Filtering:  $y(n) = w(0)x(n) + w(1)x(n-1)$

Output:  $e(n) = d(n) - y(n)$

$$w(0) = w(0) + 2\mu e(n)x(n)$$

$$w(1) = w(1) + 2\mu e(n)x(n-1)$$

# Noise Canceller Using Adaptive Filter-2

For  $n = 0$

Digital filtering:

$$y(0) = w(0)x(0) + w(1)x(-1) = 0 \times 1 + 0 \times 0 = 0$$

Computing the output:

$$e(0) = d(0) - y(0) = 2 - 0 = 2$$

Updating coefficients:

$$w(0) = w(0) + 2 \times 0.1 \times e(0)x(0) = 0 + 2 \times 0.1 \times 2 \times 1 = 0.4$$

$$w(1) = w(1) + 2 \times 0.1 \times e(0)x(-1) = 0 + 2 \times 0.1 \times 2 \times 0 = 0.0$$

---

For  $n = 1$

Digital filtering:

$$y(1) = w(0)x(1) + w(1)x(0) = 0.4 \times 1 + 0 \times 1 = 0.4$$

Computing the output:

$$e(1) = d(1) - y(1) = 1 - 0.4 = 0.6$$

Updating coefficients:

$$w(0) = w(0) + 2 \times 0.1 \times e(1)x(1) = 0.4 + 2 \times 0.1 \times 0.6 \times 1 = 0.52$$

$$w(1) = w(1) + 2 \times 0.1 \times e(1)x(0) = 0 + 2 \times 0.1 \times 0.6 \times 1 = 0.12$$

# Noise Canceller Using Adaptive Filter-3

For  $n = 2$

Digital filtering:

$$y(2) = w(0)x(2) + w(1)x(1) = 0.52 \times (-1) + 0.12 \times 1 = -0.4$$

Computing the output:

$$e(2) = d(2) - y(2) = -2 - (-0.4) = -1.6$$

Updating coefficients:

$$w(0) = w(0) + 2 \times 0.1 \times e(2)x(2) = 0.52 + 2 \times 0.1 \times (-1.6) \times (-1) = 0.84$$

$$w(1) = w(1) + 2 \times 0.1 \times e(2)x(1) = 0.12 + 2 \times 0.1 \times (-1.6) \times 1 = -0.2.$$

**Output (noise-cleaned signal):**

$$e(0) = 2, e(1) = 0.6, e(2) = -1.6$$

Practice: Textbook by Li Tan, Chapter 10.

**10.3, 10.5, 10.7**

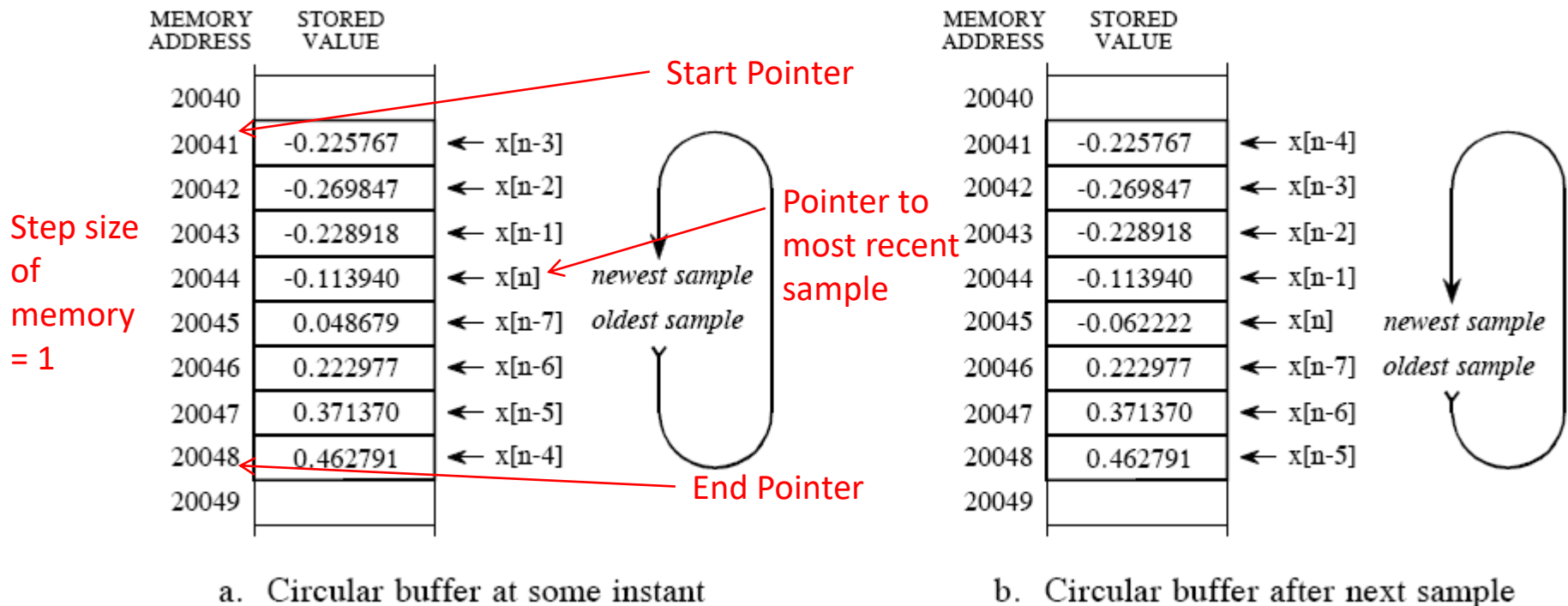
# Digital Signal Processors - Introduction

## Processors dedicated to DSP

**Off-line processing:** all the data needs to be in the memory at the same time. The output is produced after all the input data is in the memory.

**On-line processing:** the output is produced as the same time the input is coming. No delay or little delay.

## Circular Buffer Operation:

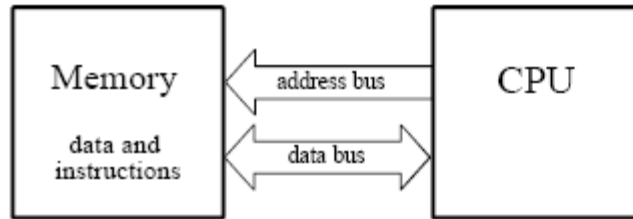


# Microprocessors Architecture

Normal microprocessors



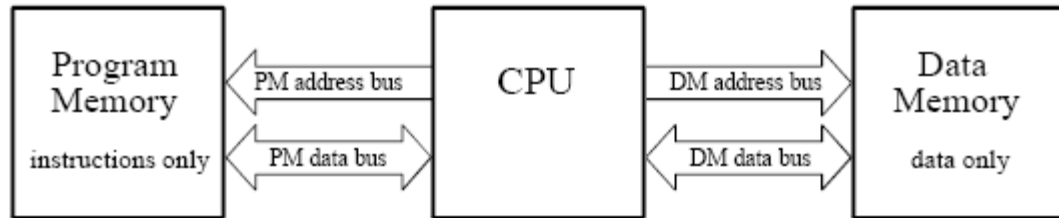
a. Von Neumann Architecture (*single memory*)



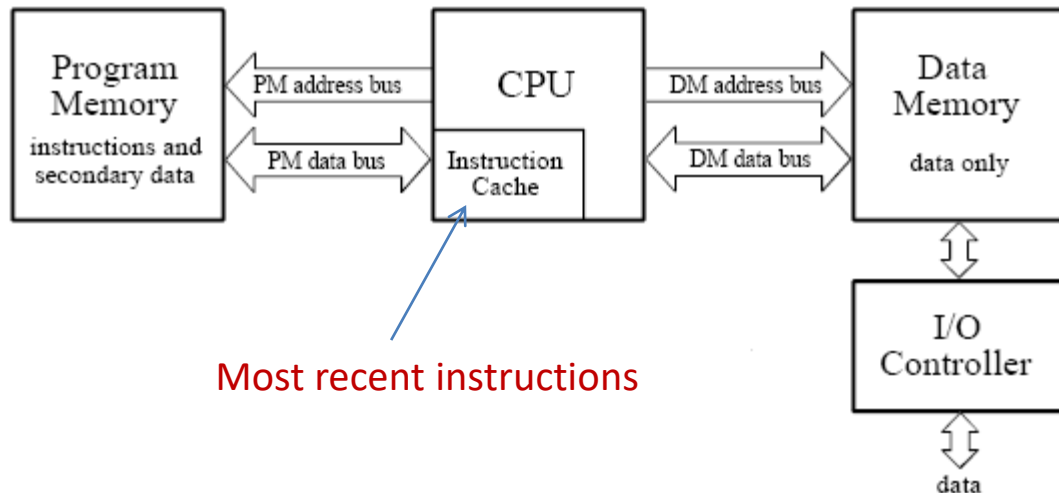
DSP uses these ideas



b. Harvard Architecture (*dual memory*)

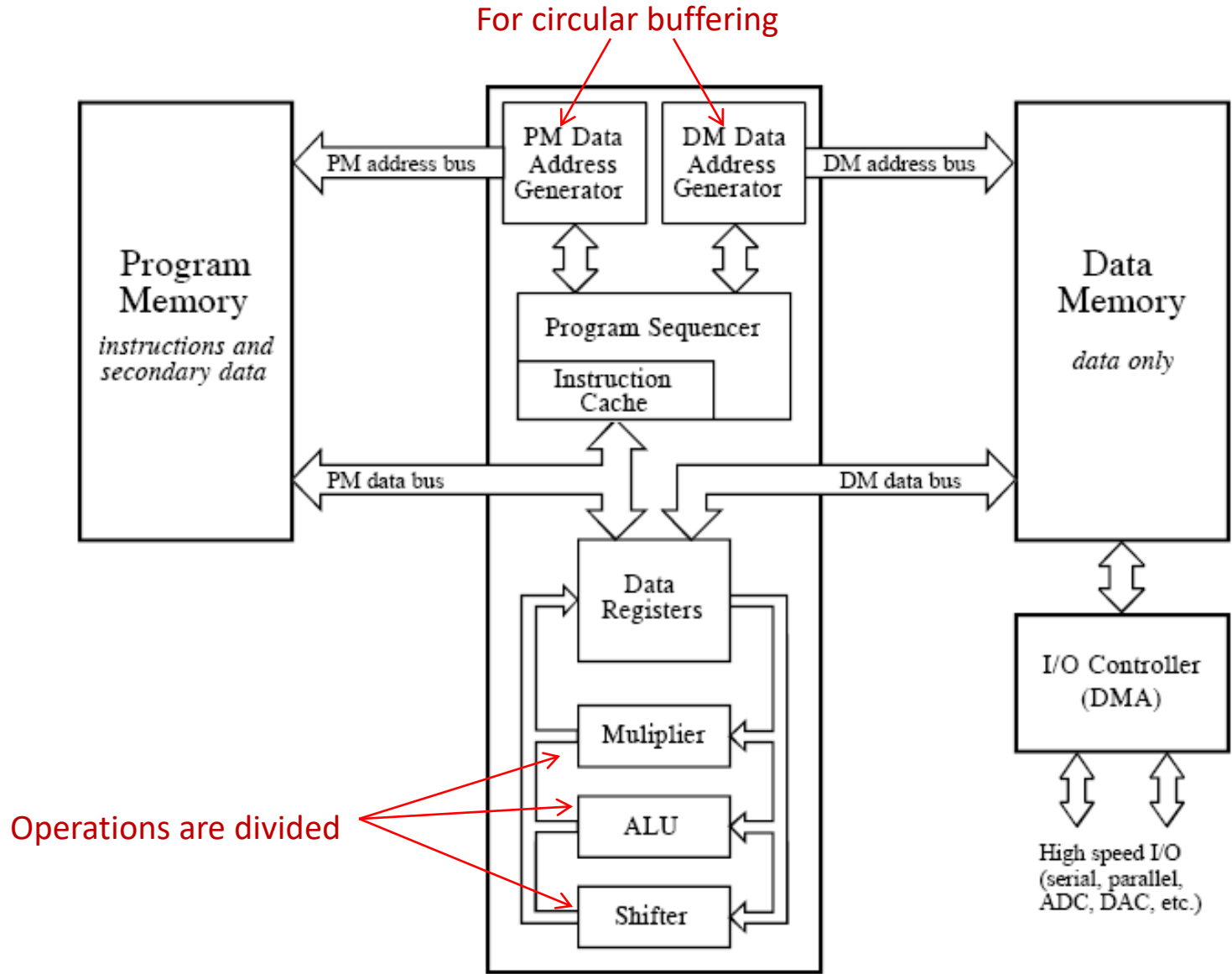


c. Super Harvard Architecture (*dual memory, instruction cache, I/O controller*)



Most recent instructions

# Typical DSP Architecture



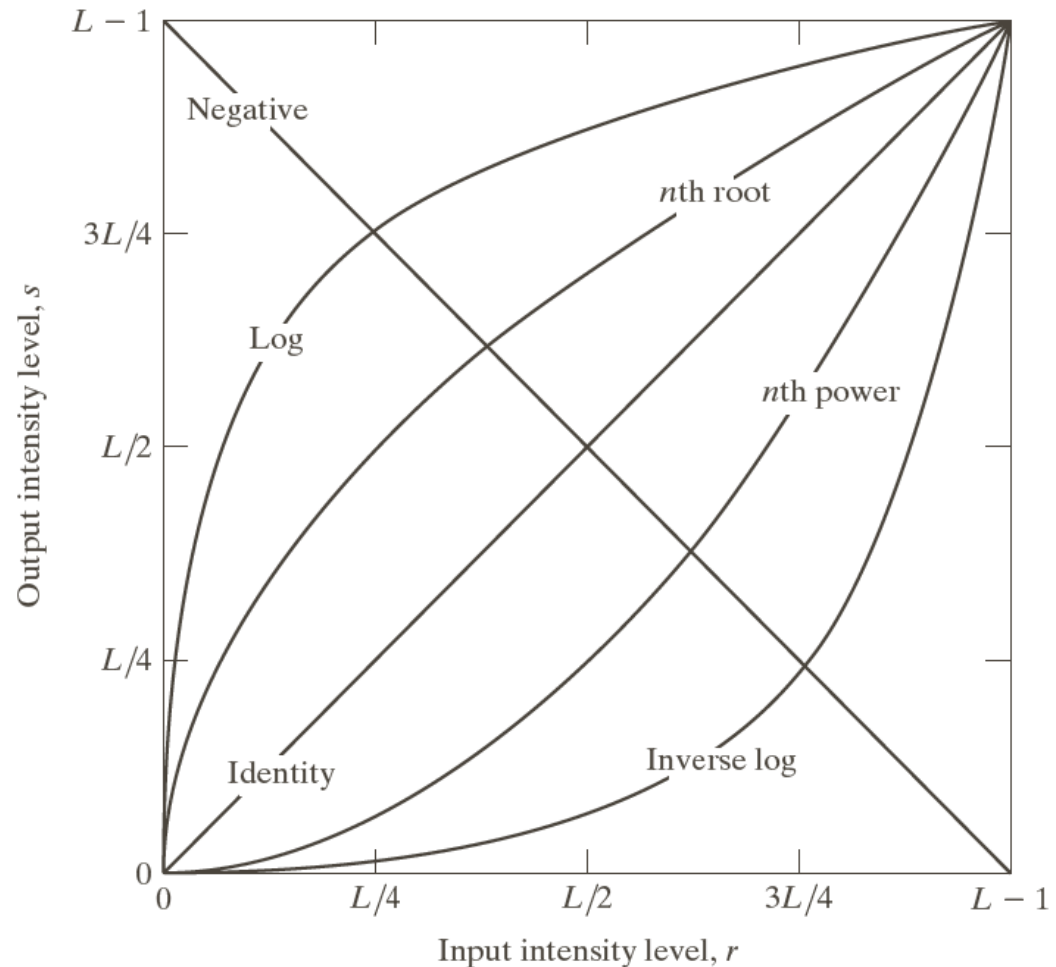


# Application of DSP in Image Processing: Basic Intensity Transfer Functions

- Linear
- Logarithmic
- Power-law

Grey Image has intensity in the range  $[0 - 255]$  i.e. it uses 8 bits.

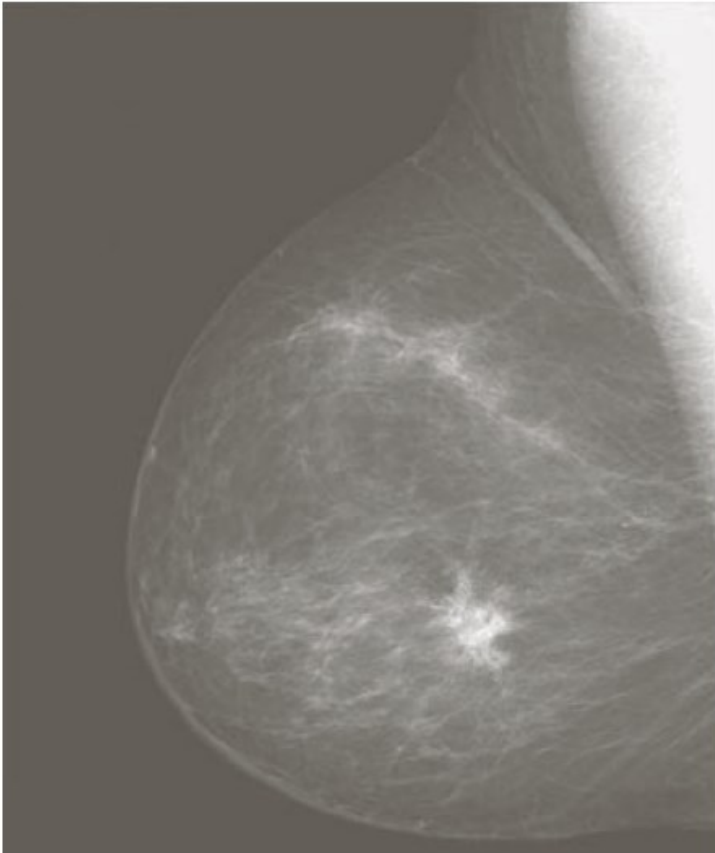
Intensity is transformed for better viewing.



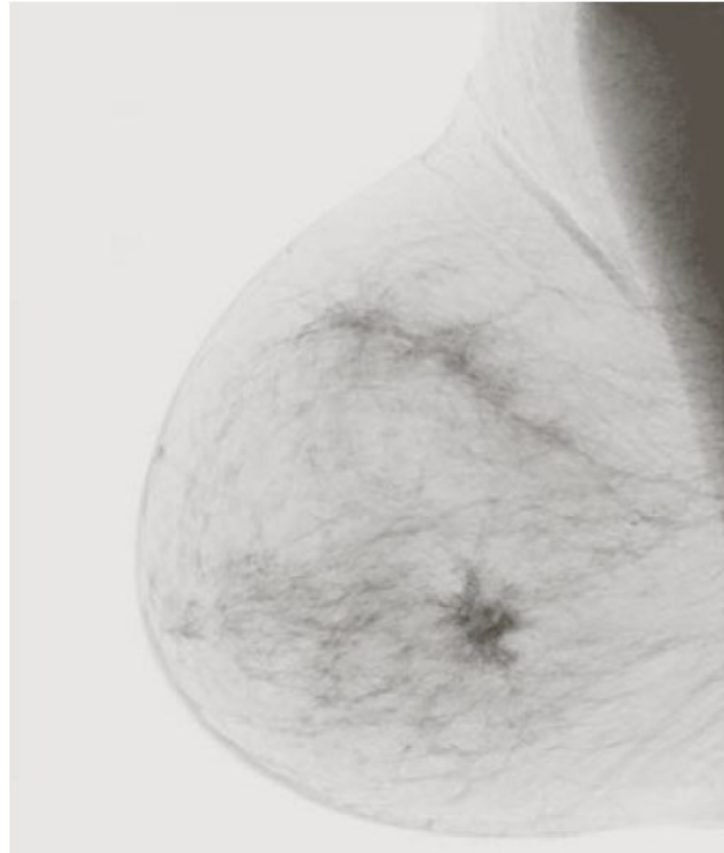
# Negative Transformation

Intensity level:  $[0, L - 1]$

$$s = L - 1 - r$$



Original Image



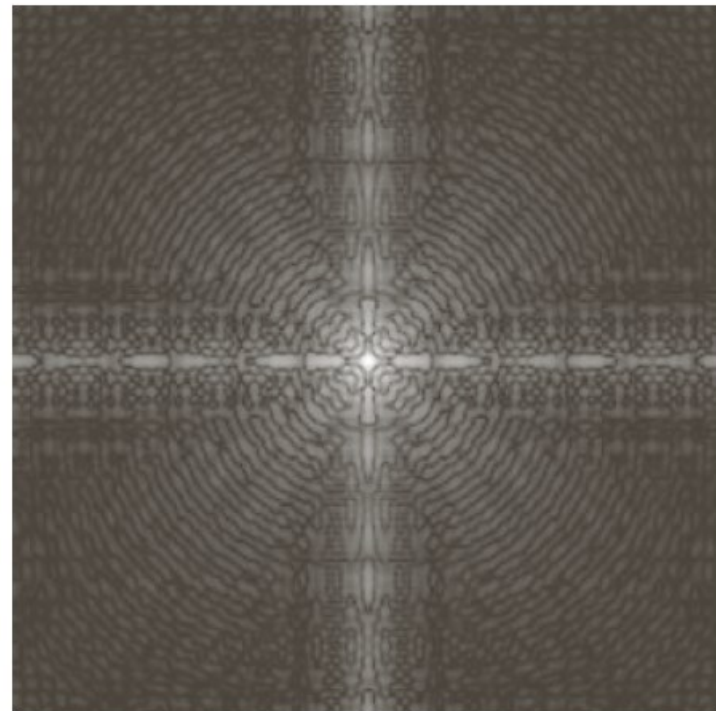
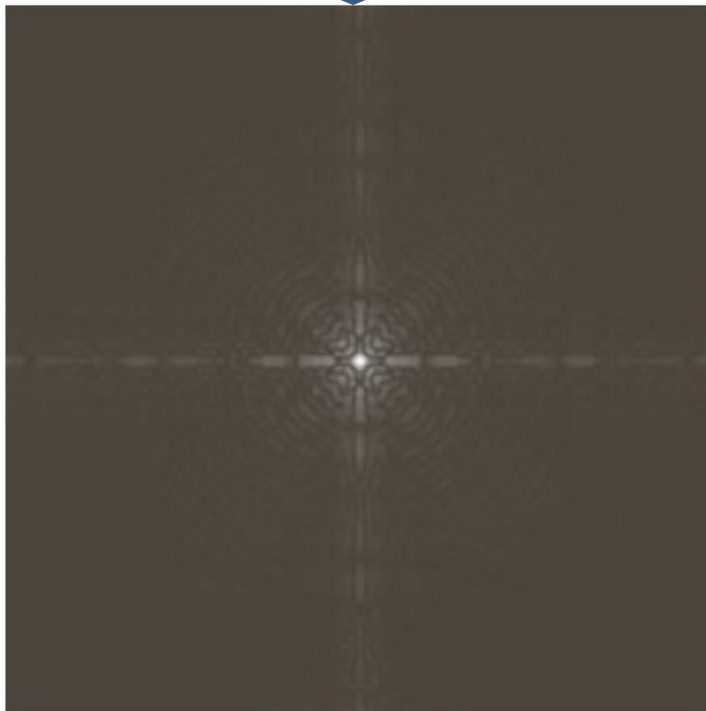
Negative Image

# Log Transformation

$$s = c \log(1 + r)$$

Compresses the dynamic range of images with large variations in pixel values.

Loss of details in low pixel values



Original Image (Fourier Spectrum)

Log Transformed Image

# Contrast manipulation: Power-Law

MRI of a fractured spine.

$$s = cr^\lambda$$



$\lambda = 0.6$



$\lambda = 0.4$

Best contrast



$\lambda = 0.3$

Washed out

# Filter Mask

$w_1$	$w_2$	$w_3$
$w_4$	$w_5$	$w_6$
$w_7$	$w_8$	$w_9$

$$R = w_1 z_1 + w_2 z_2 + \dots + w_9 z_9 = \sum_{k=1}^9 w_k z_k = \mathbf{w}^T \mathbf{z}$$

## Smoothing Filter (low pass) mask:

 $\frac{1}{9} \times$ 

1	1	1
1	1	1
1	1	1

↑  
Equal weight

 $\frac{1}{16} \times$ 

1	2	1
2	4	2
1	2	1

↑  
Weighted average

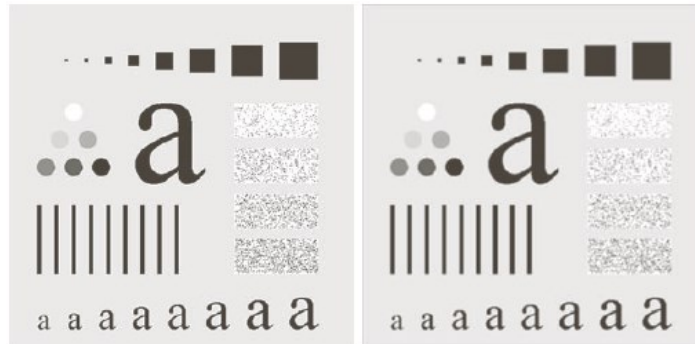
$$g(x, y) = \frac{\sum_{s=-a}^a \sum_{t=-b}^b w(s, t) f(x+s, y+t)}{\sum_{s=-a}^a \sum_{t=-b}^b w(s, t)}$$



Normalization factor

# Smoothing Effect

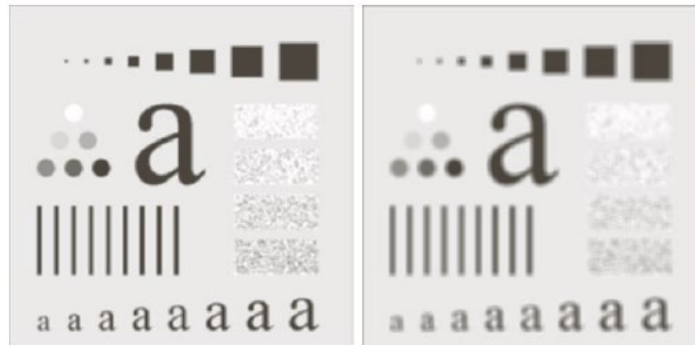
Original image



3 X 3 mask

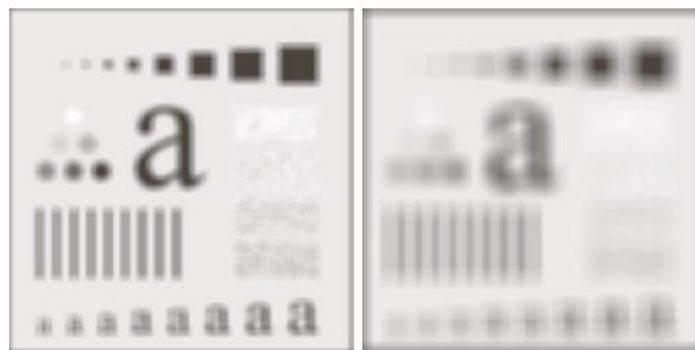
Noise is less pronounced.

5 X 5 mask



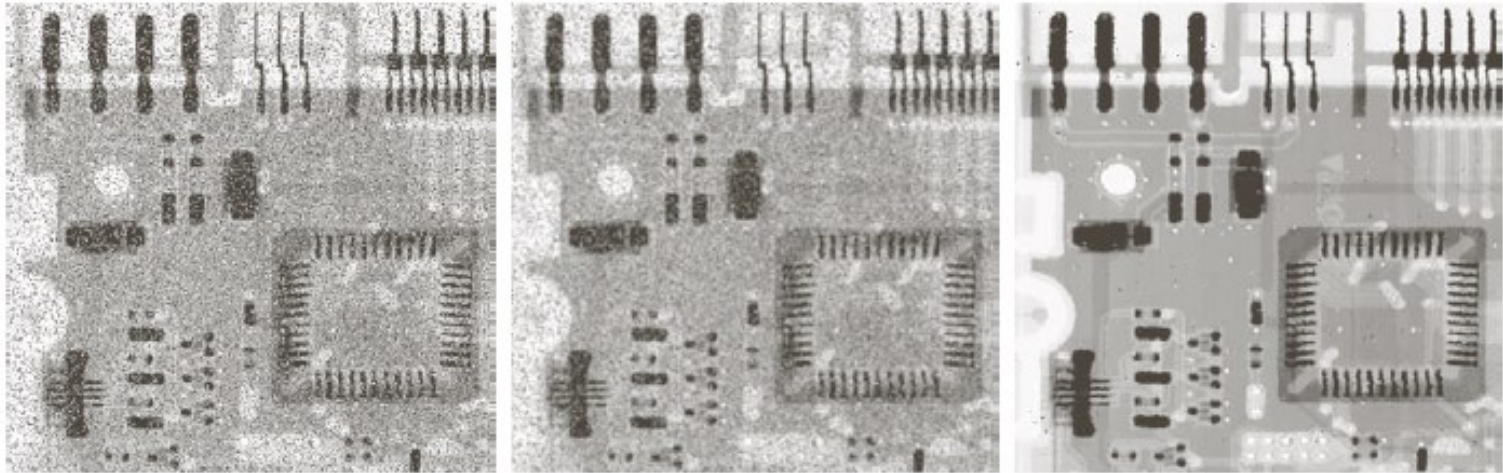
35 X 35 mask

Completely blurred!



# Median Filtering

Find the median in the neighborhood, then assign the center pixel value to that median.



a b c

**FIGURE 3.35** (a) X-ray image of circuit board corrupted by salt-and-pepper noise. (b) Noise reduction with a  $3 \times 3$  averaging mask. (c) Noise reduction with a  $3 \times 3$  median filter. (Original image courtesy of Mr. Joseph E. Pascente, Lixi, Inc.)

# Digital (Image) Watermarking

Inserting information (watermark) into images in such a way that the watermark is inseparable from the images.

- Copyright identification.
- User identification.
- Authenticity determination.
- Automated monitoring.
- Copy protection.

$$f_w = (1 - \alpha)f + \alpha w$$

Watermarked image

Original image

Watermark



# Example of Watermarking - I

Visible



a  
b c

**FIGURE 8.50**  
A simple visible watermark:  
(a) watermark;  
(b) the watermarked image; and (c) the difference between the watermarked image and the original (non-watermarked) image.

$\alpha = 0.3$

# Example of Watermarking - II

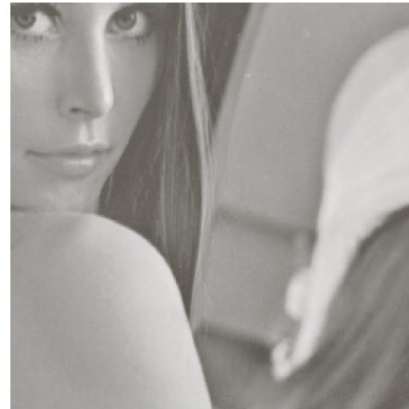
## Invisible

Watermark is inserted in image's two LSBs.

$$f_w = 4 \left( \frac{f}{4} \right) + \frac{w}{64}$$

Sets two LSBs of image to 00.

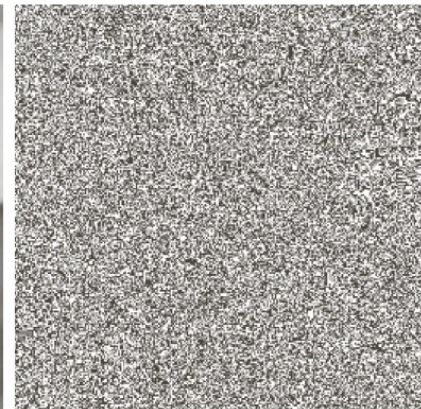
Shifts two MSBs into two LSBs.



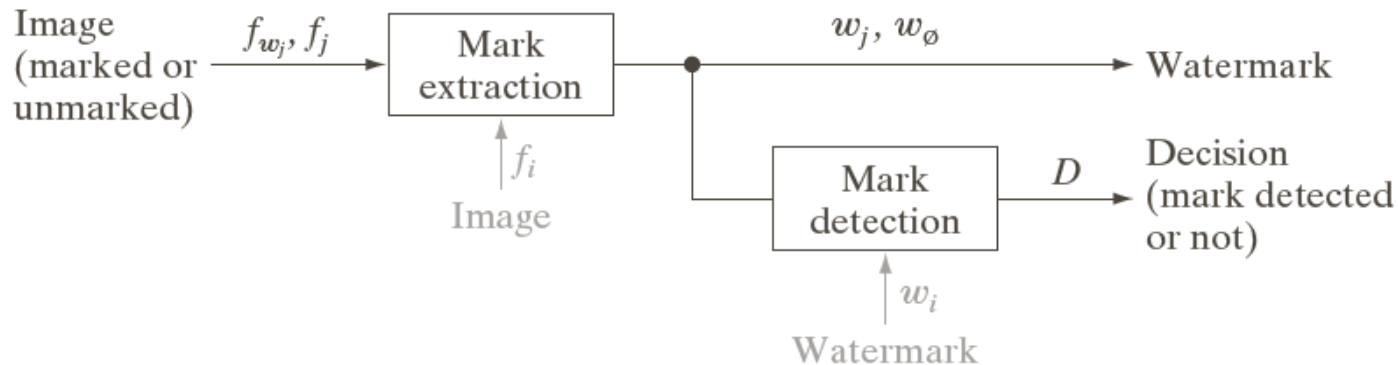
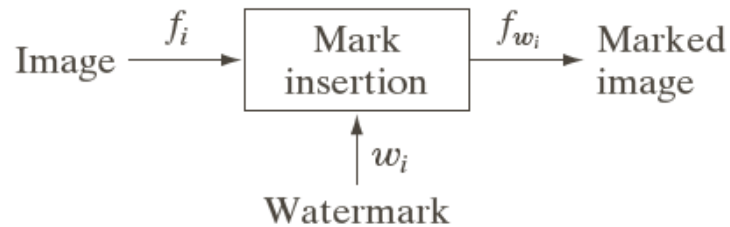
By zeroing 6 MSB and scaling the remains to full intensity level, we get



Digital Image Processing



# Image Watermarking System



a  
b

**FIGURE 8.52**

A typical image watermarking system:  
(a) encoder;  
(b) decoder.

**Private / restricted key system:**  $f_i$  and  $w_i$  are used.

**Public / unrestricted key system:**  $f_i$  and  $w_i$  are unused.